

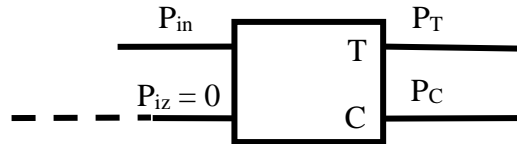
Subject no. 1

1. $z = 1.165 + j \cdot 0.720$; $Y = 1 / 50\Omega / (1.165 + j \cdot 0.720) = 0.0124S + j \cdot (-0.0077)S$; $\Gamma = (z-1)/(z+1) = (1.165 + j \cdot 0.720 - 1)/(1.165 + j \cdot 0.720 + 1) = 0.168 + j \cdot (0.277) = 0.324 \angle 58.7^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.098$, $Z_{0E} = 55.150\Omega$, $Z_{0O} = 45.331\Omega$

b) $P_c = 71.5\mu W = -11.457\text{dBm}$; $P_{in} = P_c + C = -11.457\text{dBm} + 20.2\text{dB} = 8.743\text{dBm} = 7.487 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [\text{mW}] = P_{in} - P_c - P_{iz} = 7.487\text{mW} - 0.0715\text{mW} - 0 = 7.415 \text{ mW} = 8.701 \text{ dBm}$



3. The shunt RC load with $R = 45\Omega$ and $C = 0.364\text{pF}$ has $Z_L = 23.01\Omega + j \cdot (-22.49)\Omega$ at 9.5GHz

$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$ (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 3/5) = 0.727$; $\cot(\beta l) = 1.376$; $Z_{in} = 23.97\Omega \angle 30.2^\circ = 20.72\Omega + j \cdot (12.05)\Omega$;

b) If the line becomes open-circuited $Z_L = \infty$; $Z_{in} = j \cdot (-82.58)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 11\text{dB} + 16\text{dB} + 10\text{dB} = 37\text{dB}$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$F_1 = 2.75\text{dB} = 1.884$, $G_1 = 11\text{dB} = 12.589$, $F_2 = 2.15\text{dB} = 1.641$, $G_2 = 16\text{dB} = 39.811$, $F_3 = 2.80\text{dB} = 1.905$,
 $F = 1.884 + (1.641 - 1)/12.589 + (1.905 - 1)/12.589/39.811 = 1.936 = 2.870\text{dB}$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ |
|----|-----------------------------|------------|-------|-------|
| T1 | $-0.369 + j \cdot (-0.328)$ | 0.493 | 0.569 | 0.446 |
| T2 | $-0.155 + j \cdot (-0.374)$ | 0.405 | 0.675 | 0.735 |

b) $\mu (T1) < \mu (T2)$ so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.316$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 31.484 = 14.981 \text{ dB}$ (L9/2023, S75)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.308$, $\arg(S_{22}^*) = 96.3^\circ$; $\theta_{S1} = 5.8^\circ$; $\text{Im}(y_S) = -0.647$; **or** $\theta_{S2} = 77.9^\circ$; $\text{Im}(y_S) = 0.647$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.749$, $\arg(S_{11}^*) = 92.0^\circ$; $\theta_{S3} = 23.3^\circ$; $\text{Im}(y_S) = -2.261$; **or** $\theta_{S4} = 64.7^\circ$; $\text{Im}(y_S) = 2.261$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.647) + (-2.261) = -2.908$; $\theta_{S1} = 5.8^\circ$; $\theta_{P1} = 109.0^\circ$; $\theta_{S3} = 23.3^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.647) + (2.261) = 1.613$; $\theta_{S1} = 5.8^\circ$; $\theta_{P2} = 58.2^\circ$; $\theta_{S4} = 64.7^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.647) + (-2.261) = -1.613$; $\theta_{S2} = 77.9^\circ$; $\theta_{P3} = 121.8^\circ$; $\theta_{S3} = 23.3^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.647) + (2.261) = 2.908$; $\theta_{S2} = 77.9^\circ$; $\theta_{P4} = 71.0^\circ$; $\theta_{S4} = 64.7^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

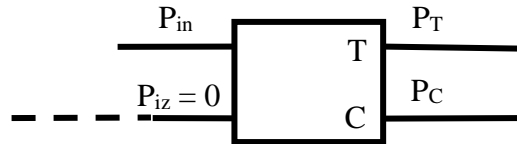
Subject no. 2

1. $z = 0.820 + j \cdot 1.170$; $Y = 1 / 50\Omega / (0.820 + j \cdot 1.170) = 0.0080S + j \cdot (-0.0115)S$; $\Gamma = (z-1)/(z+1) = (0.820 + j \cdot 1.170 - 1)/(0.820 + j \cdot 1.170 + 1) = 0.222 + j \cdot (-0.500) = 0.547 \angle 66.0^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.082$, $Z_{0E} = 54.295\Omega$, $Z_{0O} = 46.045\Omega$

b) $P_c = 127.0\mu W = -8.962\text{dBm}$; $P_{in} = P_c + C = -8.962\text{dBm} + 21.7\text{dB} = 12.738\text{dBm} = 18.785 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [\text{mW}] = P_{in} - P_c - P_{iz} = 18.785\text{mW} - 0.1270\text{mW} - 0 = 18.658 \text{ mW} = 12.709 \text{ dBm}$



3. The shunt RC load with $R = 30\Omega$ and $C = 0.408\text{pF}$ has $Z_L = 21.32\Omega + j \cdot (-13.61)\Omega$ at 8.3GHz

$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$ (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 3/6) = 0.000$; $\cot(\beta l) = \infty$; $Z_{in} = 25.29\Omega \angle -32.6^\circ = 21.32\Omega + j \cdot (-13.61)\Omega$;

b) If the line becomes short-circuited $Z_L = 0$; $Z_{in} = j \cdot (0.00)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 13\text{dB} + 19\text{dB} + 11\text{dB} = 43\text{dB}$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$F_1 = 2.74\text{dB} = 1.879$, $G_1 = 13\text{dB} = 19.953$, $F_2 = 2.20\text{dB} = 1.660$, $G_2 = 19\text{dB} = 79.433$, $F_3 = 2.26\text{dB} = 1.683$,
 $F = 1.879 + (1.660 - 1)/19.953 + (1.683 - 1)/19.953/79.433 = 1.913 = 2.817\text{dB}$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ |
|----|-----------------------------|------------|-------|-------|
| T1 | $-0.472 + j \cdot (0.086)$ | 0.479 | 0.753 | 0.637 |
| T2 | $-0.234 + j \cdot (-0.161)$ | 0.284 | 0.877 | 0.903 |

b) $\mu (T1) < \mu (T2)$ so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.094$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 24.045 = 13.810 \text{ dB}$ (L9/2023, S75)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.159$, $\arg(S_{22}^*) = 139.6^\circ$; $\theta_{S1} = 159.8^\circ$; $\text{Im}(y_S) = -0.322$; **or** $\theta_{S2} = 60.6^\circ$; $\text{Im}(y_S) = 0.322$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.637$, $\arg(S_{11}^*) = 127.9^\circ$; $\theta_{S3} = 0.8^\circ$; $\text{Im}(y_S) = -1.653$; **or** $\theta_{S4} = 51.3^\circ$; $\text{Im}(y_S) = 1.653$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.322) + (-1.653) = -1.975$; $\theta_{S1} = 159.8^\circ$; $\theta_{P1} = 116.9^\circ$; $\theta_{S3} = 0.8^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.322) + (1.653) = 1.331$; $\theta_{S1} = 159.8^\circ$; $\theta_{P2} = 53.1^\circ$; $\theta_{S4} = 51.3^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.322) + (-1.653) = -1.331$; $\theta_{S2} = 60.6^\circ$; $\theta_{P3} = 126.9^\circ$; $\theta_{S3} = 0.8^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.322) + (1.653) = 1.975$; $\theta_{S2} = 60.6^\circ$; $\theta_{P4} = 63.1^\circ$; $\theta_{S4} = 51.3^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

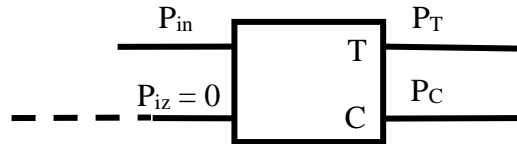
Subject no. 3

1. $z = 1.220 - j \cdot 0.755$; $Y = 1 / 50\Omega / (1.220 - j \cdot 0.755) = 0.0119S + j \cdot (0.0073)S$; $\Gamma = (z-1)/(z+1) = (1.220 - j \cdot 0.755 - 1)/(1.220 - j \cdot 0.755 + 1) = 0.192 + j \cdot (-0.275) = 0.335 \angle -55.0^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.082$, $Z_{0E} = 54.295\Omega$, $Z_{0O} = 46.045\Omega$

b) $P_c = 143.5\mu W = -8.431\text{dBm}$; $P_{in} = P_c + C = -8.431\text{dBm} + 21.7\text{dB} = 13.269\text{dBm} = 21.225 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [\text{mW}] = P_{in} - P_c - P_{iz} = 21.225\text{mW} - 0.1435\text{mW} - 0 = 21.082 \text{ mW} = 13.239 \text{ dBm}$



3. The series RL load with $R = 50\Omega$ and $L = 0.617\text{nH}$ has $Z_L = 50.00\Omega + j \cdot (32.56)\Omega$ at 8.4GHz

$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$ (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 1/3) = -1.732$; $\cot(\beta l) = -0.577$; $Z_{in} = 41.13\Omega \angle -22.5^\circ = 38.00\Omega + j \cdot (-15.74)\Omega$;

b) If the line becomes short-circuited $Z_L = 0$; $Z_{in} = j \cdot (-112.58)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 17\text{dB} + 15\text{dB} + 13\text{dB} = 45\text{dB}$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$F_1 = 2.44\text{dB} = 1.754$, $G_1 = 17\text{dB} = 50.119$, $F_2 = 2.15\text{dB} = 1.641$, $G_2 = 15\text{dB} = 31.623$, $F_3 = 2.11\text{dB} = 1.626$,
 $F = 1.754 + (1.641 - 1)/50.119 + (1.626 - 1)/50.119/31.623 = 1.767 = 2.473\text{dB}$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ' |
|----|-----------------------------|------------|-------|--------|
| T1 | $-0.477 + j \cdot (0.066)$ | 0.481 | 0.746 | 0.808 |
| T2 | $-0.234 + j \cdot (-0.173)$ | 0.291 | 0.866 | 0.936 |

b) $\mu' (T1) < \mu' (T2)$ so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.101$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 24.182 = 13.835 \text{ dB}$ (L9/2023, S75)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.167$, $\arg(S_{22}^*) = 136.1^\circ$; $\theta_{S1} = 161.8^\circ$; $\text{Im}(y_S) = -0.339$; **or** $\theta_{S2} = 62.1^\circ$; $\text{Im}(y_S) = 0.339$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.640$, $\arg(S_{11}^*) = 126.0^\circ$; $\theta_{S3} = 1.9^\circ$; $\text{Im}(y_S) = -1.666$; **or** $\theta_{S4} = 52.1^\circ$; $\text{Im}(y_S) = 1.666$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.339) + (-1.666) = -2.005$; $\theta_{S1} = 161.8^\circ$; $\theta_{P1} = 116.5^\circ$; $\theta_{S3} = 1.9^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.339) + (1.666) = 1.327$; $\theta_{S1} = 161.8^\circ$; $\theta_{P2} = 53.0^\circ$; $\theta_{S4} = 52.1^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.339) + (-1.666) = -1.327$; $\theta_{S2} = 62.1^\circ$; $\theta_{P3} = 127.0^\circ$; $\theta_{S3} = 1.9^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.339) + (1.666) = 2.005$; $\theta_{S2} = 62.1^\circ$; $\theta_{P4} = 63.5^\circ$; $\theta_{S4} = 52.1^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

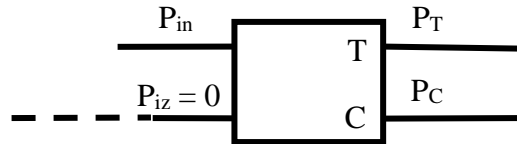
Subject no. 4

1. $z = 1.230 + j \cdot 1.105$; $Y = 1 / 50\Omega / (1.230 + j \cdot 1.105) = 0.0090S + j \cdot (-0.0081)S$; $\Gamma = (z-1)/(z+1) = (1.230 + j \cdot 1.105 - 1)/(1.230 + j \cdot 1.105 + 1) = 0.280 + j \cdot (0.357) = 0.454 \angle 51.9^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.107$, $Z_{0E} = 55.678\Omega$, $Z_{0O} = 44.901\Omega$

b) $P_c = 51.5\mu W = -12.882\text{dBm}$; $P_{in} = P_c + C = -12.882\text{dBm} + 19.4\text{dB} = 6.518\text{dBm} = 4.485 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [\text{mW}] = P_{in} - P_c - P_{iz} = 4.485\text{mW} - 0.0515\text{mW} - 0 = 4.434 \text{ mW} = 6.468 \text{ dBm}$



3. The series RC load with $R = 33\Omega$ and $C = 0.362\text{pF}$ has $Z_L = 33.00\Omega + j \cdot (-59.41)\Omega$ at 7.4GHz

$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$ (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 2/6) = -1.732$; $\cot(\beta l) = -0.577$; $Z_{in} = 116.64\Omega \angle 52.0^\circ = 71.79\Omega + j \cdot (91.92)\Omega$;

b) If the line becomes short-circuited $Z_L = 0$; $Z_{in} = j \cdot (-95.26)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 17\text{dB} + 14\text{dB} + 18\text{dB} = 49\text{dB}$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$F_1 = 2.75\text{dB} = 1.884$, $G_1 = 17\text{dB} = 50.119$, $F_2 = 2.04\text{dB} = 1.600$, $G_2 = 14\text{dB} = 25.119$, $F_3 = 2.63\text{dB} = 1.832$,
 $F = 1.884 + (1.600 - 1)/50.119 + (1.832 - 1)/50.119/25.119 = 1.896 = 2.779\text{dB}$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ |
|----|-----------------------------|------------|-------|-------|
| T1 | $-0.475 + j \cdot (-0.082)$ | 0.482 | 0.670 | 0.547 |
| T2 | $-0.225 + j \cdot (-0.237)$ | 0.326 | 0.803 | 0.844 |

b) $\mu (T1) < \mu (T2)$ so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.166$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 26.558 = 14.242 \text{ dB}$ (L9/2023, S75)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.223$, $\arg(S_{22}^*) = 121.5^\circ$; $\theta_{S1} = 170.7^\circ$; $\text{Im}(y_S) = -0.458$; **or** $\theta_{S2} = 67.8^\circ$; $\text{Im}(y_S) = 0.458$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.676$, $\arg(S_{11}^*) = 114.0^\circ$; $\theta_{S3} = 9.3^\circ$; $\text{Im}(y_S) = -1.835$; **or** $\theta_{S4} = 56.7^\circ$; $\text{Im}(y_S) = 1.835$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.458) + (-1.835) = -2.292$; $\theta_{S1} = 170.7^\circ$; $\theta_{P1} = 113.6^\circ$; $\theta_{S3} = 9.3^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.458) + (1.835) = 1.377$; $\theta_{S1} = 170.7^\circ$; $\theta_{P2} = 54.0^\circ$; $\theta_{S4} = 56.7^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.458) + (-1.835) = -1.377$; $\theta_{S2} = 67.8^\circ$; $\theta_{P3} = 126.0^\circ$; $\theta_{S3} = 9.3^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.458) + (1.835) = 2.292$; $\theta_{S2} = 67.8^\circ$; $\theta_{P4} = 66.4^\circ$; $\theta_{S4} = 56.7^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

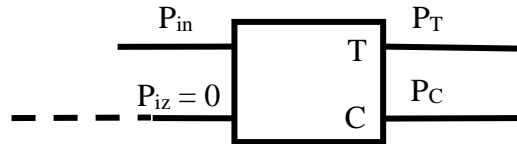
Subject no. 5

1. $z = 1.005 - j \cdot 1.080$; $Y = 1 / 50\Omega / (1.005 - j \cdot 1.080) = 0.0092S + j \cdot (0.0099)S$; $\Gamma = (z-1)/(z+1) = (1.005 - j \cdot 1.080 - 1)/(1.005 - j \cdot 1.080 + 1) = 0.227 + j \cdot (-0.416) = 0.474 \angle -61.4^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.095$, $Z_{0E} = 55.026\Omega$, $Z_{0O} = 45.433\Omega$

b) $P_c = 126.0\mu W = -8.996\text{dBm}$; $P_{in} = P_c + C = -8.996\text{dBm} + 20.4\text{dB} = 11.404\text{dBm} = 13.816 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [\text{mW}] = P_{in} - P_c - P_{iz} = 13.816\text{mW} - 0.1260\text{mW} - 0 = 13.690 \text{ mW} = 11.364 \text{ dBm}$



3. The shunt RL load with $R = 52\Omega$ and $L = 0.612\text{nH}$ has $Z_L = 13.98\Omega + j \cdot (23.05)\Omega$ at 8.2GHz

$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$ (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 5/8) = 1.000$; $\cot(\beta l) = 1.000$; $Z_{in} = 134.47\Omega \angle 64.9^\circ = 57.10\Omega + j \cdot (121.75)\Omega$;

b) If the line becomes short-circuited $Z_L = 0$; $Z_{in} = j \cdot (70.00)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 11\text{dB} + 15\text{dB} + 17\text{dB} = 43\text{dB}$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$F_1 = 2.08\text{dB} = 1.614$, $G_1 = 11\text{dB} = 12.589$, $F_2 = 2.46\text{dB} = 1.762$, $G_2 = 15\text{dB} = 31.623$, $F_3 = 2.84\text{dB} = 1.923$,
 $F = 1.614 + (1.762 - 1)/12.589 + (1.923 - 1)/12.589/31.623 = 1.677 = 2.246\text{dB}$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ |
|----|-----------------------------|------------|-------|-------|
| T1 | $-0.446 + j \cdot (0.165)$ | 0.475 | 0.788 | 0.681 |
| T2 | $-0.228 + j \cdot (-0.117)$ | 0.256 | 0.920 | 0.938 |

b) $\mu (T1) < \mu (T2)$ so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.068$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 23.500 = 13.711 \text{ dB}$ (L9/2023, S75)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.127$, $\arg(S_{22}^*) = 153.7^\circ$; $\theta_{S1} = 151.8^\circ$; $\text{Im}(y_S) = -0.256$; **or**
 $\theta_{S2} = 54.5^\circ$; $\text{Im}(y_S) = 0.256$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.625$, $\arg(S_{11}^*) = 135.5^\circ$; $\theta_{S3} = 176.6^\circ$; $\text{Im}(y_S) = -1.601$; **or**
 $\theta_{S4} = 47.9^\circ$; $\text{Im}(y_S) = 1.601$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.256) + (-1.601) = -1.857$; $\theta_{S1} = 151.8^\circ$; $\theta_{P1} = 118.3^\circ$; $\theta_{S3} = 176.6^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.256) + (1.601) = 1.345$; $\theta_{S1} = 151.8^\circ$; $\theta_{P2} = 53.4^\circ$; $\theta_{S4} = 47.9^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.256) + (-1.601) = -1.345$; $\theta_{S2} = 54.5^\circ$; $\theta_{P3} = 126.6^\circ$; $\theta_{S3} = 176.6^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.256) + (1.601) = 1.857$; $\theta_{S2} = 54.5^\circ$; $\theta_{P4} = 61.7^\circ$; $\theta_{S4} = 47.9^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

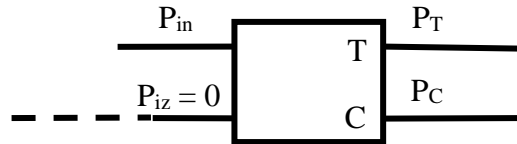
Subject no. 6

1. $z = 0.865 + j \cdot 0.795$; $Y = 1 / 50\Omega / (0.865 + j \cdot 0.795) = 0.0125S + j \cdot (-0.0115)S$; $\Gamma = (z-1)/(z+1) = (0.865 + j \cdot 0.795 - 1)/(0.865 + j \cdot 0.795 + 1) = 0.093 + j \cdot (0.387) = 0.398 \angle 76.6^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.097$, $Z_{0E} = 55.088\Omega$, $Z_{0O} = 45.382\Omega$

b) $P_c = 78.0\mu W = -11.079\text{dBm}$; $P_{in} = P_c + C = -11.079\text{dBm} + 20.3\text{dB} = 9.221\text{dBm} = 8.358 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [\text{mW}] = P_{in} - P_c - P_{iz} = 8.358\text{mW} - 0.0780\text{mW} - 0 = 8.280 \text{ mW} = 9.180 \text{ dBm}$



3. The shunt RL load with $R = 35\Omega$ and $L = 0.733\text{nH}$ has $Z_L = 21.86\Omega + j \cdot (16.95)\Omega$ at 9.8GHz

$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$ (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 2/3) = 1.732$; $\cot(\beta l) = 0.577$; $Z_{in} = 219.32\Omega \angle 50.8^\circ = 138.56\Omega + j \cdot (170.00)\Omega$;

b) If the line becomes short-circuited $Z_L = 0$; $Z_{in} = j \cdot (155.88)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 12\text{dB} + 17\text{dB} + 18\text{dB} = 47\text{dB}$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$F_1 = 2.99\text{dB} = 1.991$, $G_1 = 12\text{dB} = 15.849$, $F_2 = 2.18\text{dB} = 1.652$, $G_2 = 17\text{dB} = 50.119$, $F_3 = 2.17\text{dB} = 1.648$,
 $F = 1.991 + (1.652 - 1)/15.849 + (1.648 - 1)/15.849/50.119 = 2.033 = 3.081\text{dB}$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ' |
|----|-----------------------------|------------|-------|--------|
| T1 | $-0.366 + j \cdot (0.297)$ | 0.471 | 0.846 | 0.872 |
| T2 | $-0.204 + j \cdot (-0.056)$ | 0.211 | 1.005 | 1.002 |

b) $\mu'(T1) < \mu'(T2)$ so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.039$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 22.564 = 13.534 \text{ dB}$ (L9/2023, S75); However $K = 1.005 > 1$ so maximum gain is $MAG = |S_{21}| / |S_{12}| \cdot [K - \sqrt{K^2 - 1}] = 20.410 = 13.098 \text{ dB}$ (L9/2023, S76)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.086$, $\arg(S_{22}^*) = -169.5^\circ$; $\theta_{S1} = 132.2^\circ$; $\text{Im}(y_S) = -0.173$; **or**
 $\theta_{S2} = 37.3^\circ$; $\text{Im}(y_S) = 0.173$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.604$, $\arg(S_{11}^*) = 149.0^\circ$; $\theta_{S3} = 169.1^\circ$; $\text{Im}(y_S) = -1.516$; **or**
 $\theta_{S4} = 41.9^\circ$; $\text{Im}(y_S) = 1.516$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.173) + (-1.516) = -1.688$; $\theta_{S1} = 132.2^\circ$; $\theta_{P1} = 120.6^\circ$; $\theta_{S3} = 169.1^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.173) + (1.516) = 1.343$; $\theta_{S1} = 132.2^\circ$; $\theta_{P2} = 53.3^\circ$; $\theta_{S4} = 41.9^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.173) + (-1.516) = -1.343$; $\theta_{S2} = 37.3^\circ$; $\theta_{P3} = 126.7^\circ$; $\theta_{S3} = 169.1^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.173) + (1.516) = 1.688$; $\theta_{S2} = 37.3^\circ$; $\theta_{P4} = 59.4^\circ$; $\theta_{S4} = 41.9^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

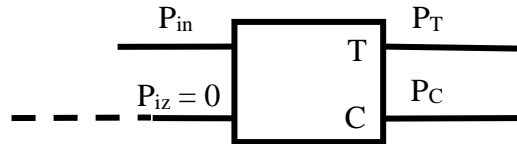
Subject no. 7

1. $z = 0.790 - j \cdot 1.270$; $Y = 1 / 50\Omega / (0.790 - j \cdot 1.270) = 0.0071S + j \cdot (0.0114)S$; $\Gamma = (z-1)/(z+1) = (0.790 - j \cdot 1.270 - 1)/(0.790 - j \cdot 1.270 + 1) = 0.257 + j \cdot (-0.527) = 0.587 \angle -64.0^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.126$, $Z_{0E} = 56.746\Omega$, $Z_{0O} = 44.056\Omega$

b) $P_c = 111.0\mu W = -9.547\text{dBm}$; $P_{in} = P_c + C = -9.547\text{dBm} + 18.0\text{dB} = 8.453\text{dBm} = 7.004 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [\text{mW}] = P_{in} - P_c - P_{iz} = 7.004\text{mW} - 0.1110\text{mW} - 0 = 6.893 \text{ mW} = 8.384 \text{ dBm}$



3. The series RL load with $R = 67\Omega$ and $L = 1.088\text{nH}$ has $Z_L = 67.00\Omega + j \cdot (53.32)\Omega$ at 7.8GHz

$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$ (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 1/3) = -1.732$; $\cot(\beta l) = -0.577$; $Z_{in} = 15.65\Omega \angle 27.8^\circ = 13.84\Omega + j \cdot (7.31)\Omega$;

b) If the line becomes short-circuited $Z_L = 0$; $Z_{in} = j \cdot (-69.28)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 15\text{dB} + 14\text{dB} + 11\text{dB} = 40\text{dB}$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$F_1 = 2.23\text{dB} = 1.671$, $G_1 = 15\text{dB} = 31.623$, $F_2 = 2.01\text{dB} = 1.589$, $G_2 = 14\text{dB} = 25.119$, $F_3 = 2.74\text{dB} = 1.879$,
 $F = 1.671 + (1.589 - 1)/31.623 + (1.879 - 1)/31.623/25.119 = 1.691 = 2.281\text{dB}$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ |
|----|-----------------------------|------------|-------|-------|
| T1 | $-0.423 + j \cdot (-0.243)$ | 0.488 | 0.599 | 0.475 |
| T2 | $-0.190 + j \cdot (-0.322)$ | 0.374 | 0.725 | 0.779 |

b) $\mu(T1) < \mu(T2)$ so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.260$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 29.608 = 14.714 \text{ dB}$ (L9/2023, S75)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.281$, $\arg(S_{22}^*) = 105.1^\circ$; $\theta_{S1} = 0.6^\circ$; $\text{Im}(y_S) = -0.586$; **or** $\theta_{S2} = 74.3^\circ$; $\text{Im}(y_S) = 0.586$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.721$, $\arg(S_{11}^*) = 100.0^\circ$; $\theta_{S3} = 18.1^\circ$; $\text{Im}(y_S) = -2.081$; **or** $\theta_{S4} = 61.9^\circ$; $\text{Im}(y_S) = 2.081$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.586) + (-2.081) = -2.667$; $\theta_{S1} = 0.6^\circ$; $\theta_{P1} = 110.6^\circ$; $\theta_{S3} = 18.1^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.586) + (2.081) = 1.495$; $\theta_{S1} = 0.6^\circ$; $\theta_{P2} = 56.2^\circ$; $\theta_{S4} = 61.9^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.586) + (-2.081) = -1.495$; $\theta_{S2} = 74.3^\circ$; $\theta_{P3} = 123.8^\circ$; $\theta_{S3} = 18.1^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.586) + (2.081) = 2.667$; $\theta_{S2} = 74.3^\circ$; $\theta_{P4} = 69.4^\circ$; $\theta_{S4} = 61.9^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

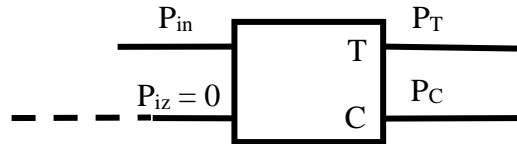
Subject no. 8

1. $z = 1.060 - j \cdot 0.990$; $Y = 1 / 50\Omega / (1.060 - j \cdot 0.990) = 0.0101S + j \cdot (0.0094)S$; $\Gamma = (z-1)/(z+1) = (1.060 - j \cdot 0.990 - 1)/(1.060 - j \cdot 0.990 + 1) = 0.211 + j \cdot (-0.379) = 0.434 \angle -60.9^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.114$, $Z_{0E} = 56.037\Omega$, $Z_{0O} = 44.613\Omega$

b) $P_c = 100.5\mu W = -9.978\text{dBm}$; $P_{in} = P_c + C = -9.978\text{dBm} + 18.9\text{dB} = 8.922\text{dBm} = 7.801 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [\text{mW}] = P_{in} - P_c - P_{iz} = 7.801\text{mW} - 0.1005\text{mW} - 0 = 7.701 \text{ mW} = 8.865 \text{ dBm}$



3. The series RC load with $R = 58\Omega$ and $C = 0.518\text{pF}$ has $Z_L = 58.00\Omega + j \cdot (-46.55)\Omega$ at 6.6GHz

$$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$$
 (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 1/3) = -1.732$; $\cot(\beta l) = -0.577$; $Z_{in} = 138.45\Omega \angle 21.4^\circ = 128.91\Omega + j \cdot (50.53)\Omega$;

b) If the line becomes short-circuited $Z_L = 0$; $Z_{in} = j \cdot (-129.90)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 12\text{dB} + 17\text{dB} + 12\text{dB} = 41\text{dB}$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$F_1 = 2.48\text{dB} = 1.770$, $G_1 = 12\text{dB} = 15.849$, $F_2 = 2.27\text{dB} = 1.687$, $G_2 = 17\text{dB} = 50.119$, $F_3 = 2.55\text{dB} = 1.799$,
 $F = 1.770 + (1.687 - 1)/15.849 + (1.799 - 1)/15.849/50.119 = 1.814 = 2.587\text{dB}$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ |
|----|----------------------------|------------|-------|-------|
| T1 | $-0.036 + j \cdot (0.489)$ | 0.490 | 0.941 | 0.898 |
| T2 | $-0.118 + j \cdot (0.007)$ | 0.118 | 1.203 | 1.150 |

b) $\mu (T1) < \mu (T2)$ so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.067$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 20.382 = 13.092 \text{ dB}$ (L9/2023, S75); However $K = 1.203 > 1$ so maximum gain is $MAG = |S_{21}| / |S_{12}| \cdot [K - \sqrt{K^2 - 1}] = 10.896 = 10.373 \text{ dB}$ (L9/2023, S76)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.169$, $\arg(S_{22}^*) = -59.3^\circ$; $\theta_{S1} = 79.5^\circ$; $\text{Im}(y_S) = -0.343$; **or**
 $\theta_{S2} = 159.8^\circ$; $\text{Im}(y_S) = 0.343$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.572$, $\arg(S_{11}^*) = -179.0^\circ$; $\theta_{S3} = 151.9^\circ$; $\text{Im}(y_S) = -1.395$; **or**
 $\theta_{S4} = 27.1^\circ$; $\text{Im}(y_S) = 1.395$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.343) + (-1.395) = -1.738$; $\theta_{S1} = 79.5^\circ$; $\theta_{P1} = 119.9^\circ$; $\theta_{S3} = 151.9^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.343) + (1.395) = 1.052$; $\theta_{S1} = 79.5^\circ$; $\theta_{P2} = 46.4^\circ$; $\theta_{S4} = 27.1^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.343) + (-1.395) = -1.052$; $\theta_{S2} = 159.8^\circ$; $\theta_{P3} = 133.6^\circ$; $\theta_{S3} = 151.9^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.343) + (1.395) = 1.738$; $\theta_{S2} = 159.8^\circ$; $\theta_{P4} = 60.1^\circ$; $\theta_{S4} = 27.1^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

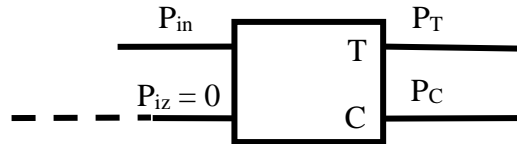
Subject no. 9

1. $z = 1.125 + j \cdot 0.935$; $Y = 1 / 50\Omega / (1.125 + j \cdot 0.935) = 0.0105S + j \cdot (-0.0087)S$; $\Gamma = (z-1)/(z+1) = (1.125 + j \cdot 0.935 - 1)/(1.125 + j \cdot 0.935 + 1) = 0.211 + j \cdot (0.347) = 0.406 \angle 58.6^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.135$, $Z_{0E} = 57.268\Omega$, $Z_{0O} = 43.654\Omega$

b) $P_c = 148.0\mu W = -8.297\text{dBm}$; $P_{in} = P_c + C = -8.297\text{dBm} + 17.4\text{dB} = 9.103\text{dBm} = 8.133 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [\text{mW}] = P_{in} - P_c - P_{iz} = 8.133\text{mW} - 0.1480\text{mW} - 0 = 7.985 \text{ mW} = 9.023 \text{ dBm}$



3. The series RL load with $R = 42\Omega$ and $L = 1.187\text{nH}$ has $Z_L = 42.00\Omega + j \cdot (71.60)\Omega$ at 9.6GHz

$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$ (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 2/3) = 1.732$; $\cot(\beta l) = 0.577$; $Z_{in} = 94.38\Omega \angle -57.6^\circ = 50.54\Omega + j \cdot (-79.70)\Omega$;

b) If the line becomes short-circuited $Z_L = 0$; $Z_{in} = j \cdot (95.26)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 12\text{dB} + 12\text{dB} + 18\text{dB} = 42\text{dB}$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$F_1 = 2.84\text{dB} = 1.923$, $G_1 = 12\text{dB} = 15.849$, $F_2 = 2.65\text{dB} = 1.841$, $G_2 = 12\text{dB} = 15.849$, $F_3 = 2.45\text{dB} = 1.758$,
 $F = 1.923 + (1.841 - 1)/15.849 + (1.758 - 1)/15.849/15.849 = 1.979 = 2.965\text{dB}$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ' |
|----|-----------------------------|------------|-------|--------|
| T1 | $-0.269 + j \cdot (0.399)$ | 0.481 | 0.881 | 0.896 |
| T2 | $-0.174 + j \cdot (-0.022)$ | 0.175 | 1.086 | 1.039 |

b) $\mu'(T1) < \mu'(T2)$ so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.037$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 21.800 = 13.385 \text{ dB}$ (L9/2023, S75); However $K = 1.086 > 1$ so maximum gain is $MAG = |S_{21}| / |S_{12}| \cdot [K - \sqrt{K^2 - 1}] = 14.435 = 11.594 \text{ dB}$ (L9/2023, S76)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.088$, $\arg(S_{22}^*) = -112.1^\circ$; $\theta_{S1} = 103.6^\circ$; $\text{Im}(y_S) = -0.177$; **or**
 $\theta_{S2} = 8.5^\circ$; $\text{Im}(y_S) = 0.177$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.586$, $\arg(S_{11}^*) = 161.0^\circ$; $\theta_{S3} = 162.4^\circ$; $\text{Im}(y_S) = -1.446$; **or**
 $\theta_{S4} = 36.6^\circ$; $\text{Im}(y_S) = 1.446$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.177) + (-1.446) = -1.623$; $\theta_{S1} = 103.6^\circ$; $\theta_{P1} = 121.6^\circ$; $\theta_{S3} = 162.4^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.177) + (1.446) = 1.270$; $\theta_{S1} = 103.6^\circ$; $\theta_{P2} = 51.8^\circ$; $\theta_{S4} = 36.6^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.177) + (-1.446) = -1.270$; $\theta_{S2} = 8.5^\circ$; $\theta_{P3} = 128.2^\circ$; $\theta_{S3} = 162.4^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.177) + (1.446) = 1.623$; $\theta_{S2} = 8.5^\circ$; $\theta_{P4} = 58.4^\circ$; $\theta_{S4} = 36.6^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

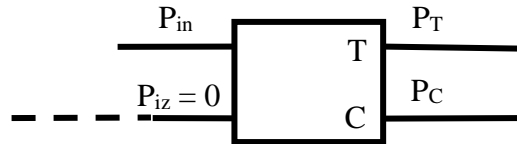
Subject no. 10

1. $z = 1.180 - j \cdot 0.805$; $Y = 1 / 50\Omega / (1.180 - j \cdot 0.805) = 0.0116S + j \cdot (0.0079)S$; $\Gamma = (z-1)/(z+1) = (1.180 - j \cdot 0.805 - 1)/(1.180 - j \cdot 0.805 + 1) = 0.193 + j \cdot (-0.298) = 0.355 \angle -57.1^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.074$, $Z_{0E} = 53.855\Omega$, $Z_{0O} = 46.421\Omega$

b) $P_c = 52.0\mu W = -12.840\text{dBm}$; $P_{in} = P_c + C = -12.840\text{dBm} + 22.6\text{dB} = 9.760\text{dBm} = 9.462 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [\text{mW}] = P_{in} - P_c - P_{iz} = 9.462\text{mW} - 0.0520\text{mW} - 0 = 9.410 \text{ mW} = 9.736 \text{ dBm}$



3. The series RL load with $R = 57\Omega$ and $L = 1.180\text{nH}$ has $Z_L = 57.00\Omega + j \cdot (55.61)\Omega$ at 7.5GHz

$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$ (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 2/6) = -1.732$; $\cot(\beta l) = -0.577$; $Z_{in} = 24.25\Omega \angle -8.0^\circ = 24.01\Omega + j \cdot (-3.38)\Omega$;

b) If the line becomes short-circuited $Z_L = 0$; $Z_{in} = j \cdot (-103.92)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 10\text{dB} + 15\text{dB} + 17\text{dB} = 42\text{dB}$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$F_1 = 2.96\text{dB} = 1.977$, $G_1 = 10\text{dB} = 10.000$, $F_2 = 2.65\text{dB} = 1.841$, $G_2 = 15\text{dB} = 31.623$, $F_3 = 2.11\text{dB} = 1.626$,
 $F = 1.977 + (1.841 - 1)/10.000 + (1.626 - 1)/10.000/31.623 = 2.063 = 3.145\text{dB}$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ |
|----|-----------------------------|------------|-------|-------|
| T1 | $-0.397 + j \cdot (-0.284)$ | 0.488 | 0.584 | 0.463 |
| T2 | $-0.174 + j \cdot (-0.348)$ | 0.389 | 0.699 | 0.757 |

b) $\mu(T1) < \mu(T2)$ so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.286$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 30.526 = 14.847 \text{ dB}$ (L9/2023, S75)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.295$, $\arg(S_{22}^*) = 100.7^\circ$; $\theta_{S1} = 3.2^\circ$; $\text{Im}(y_S) = -0.617$; **or** $\theta_{S2} = 76.1^\circ$; $\text{Im}(y_S) = 0.617$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.735$, $\arg(S_{11}^*) = 96.0^\circ$; $\theta_{S3} = 20.7^\circ$; $\text{Im}(y_S) = -2.168$; **or** $\theta_{S4} = 63.3^\circ$; $\text{Im}(y_S) = 2.168$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.617) + (-2.168) = -2.785$; $\theta_{S1} = 3.2^\circ$; $\theta_{P1} = 109.7^\circ$; $\theta_{S3} = 20.7^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.617) + (2.168) = 1.550$; $\theta_{S1} = 3.2^\circ$; $\theta_{P2} = 57.2^\circ$; $\theta_{S4} = 63.3^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.617) + (-2.168) = -1.550$; $\theta_{S2} = 76.1^\circ$; $\theta_{P3} = 122.8^\circ$; $\theta_{S3} = 20.7^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.617) + (2.168) = 2.785$; $\theta_{S2} = 76.1^\circ$; $\theta_{P4} = 70.3^\circ$; $\theta_{S4} = 63.3^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

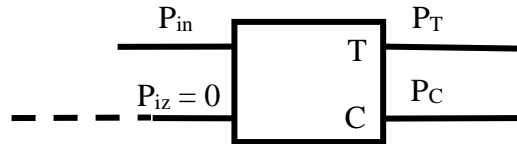
Subject no. 11

1. $z = 1.235 - j \cdot 1.250$; $Y = 1 / 50\Omega / (1.235 - j \cdot 1.250) = 0.0080S + j \cdot (0.0081)S$; $\Gamma = (z-1)/(z+1) = (1.235 - j \cdot 1.250 - 1)/(1.235 - j \cdot 1.250 + 1) = 0.318 + j \cdot (-0.381) = 0.497 \angle -50.1^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.110$, $Z_{0E} = 55.819\Omega$, $Z_{0O} = 44.788\Omega$

b) $P_c = 68.5\mu W = -11.643\text{dBm}$; $P_{in} = P_c + C = -11.643\text{dBm} + 19.2\text{dB} = 7.557\text{dBm} = 5.698 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [\text{mW}] = P_{in} - P_c - P_{iz} = 5.698\text{mW} - 0.0685\text{mW} - 0 = 5.629 \text{ mW} = 7.504 \text{ dBm}$



3. The shunt RL load with $R = 32\Omega$ and $L = 1.335\text{nH}$ has $Z_L = 23.99\Omega + j \cdot (13.86)\Omega$ at 6.6GHz

$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$ (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 2/5) = -0.727$; $\cot(\beta l) = -1.376$; $Z_{in} = 34.65\Omega \angle -41.2^\circ = 26.07\Omega + j \cdot (-22.83)\Omega$;

b) If the line becomes open-circuited $Z_L = \infty$; $Z_{in} = j \cdot (89.46)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 14\text{dB} + 14\text{dB} + 17\text{dB} = 45\text{dB}$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$F_1 = 2.37\text{dB} = 1.726$, $G_1 = 14\text{dB} = 25.119$, $F_2 = 2.86\text{dB} = 1.932$, $G_2 = 14\text{dB} = 25.119$, $F_3 = 2.30\text{dB} = 1.698$,
 $F = 1.726 + (1.932 - 1)/25.119 + (1.698 - 1)/25.119/25.119 = 1.764 = 2.465\text{dB}$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ |
|----|-----------------------------|------------|-------|-------|
| T1 | $-0.335 + j \cdot (0.330)$ | 0.471 | 0.852 | 0.768 |
| T2 | $-0.194 + j \cdot (-0.043)$ | 0.199 | 1.031 | 1.023 |

b) $\mu (T1) < \mu (T2)$ so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.039$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 22.309 = 13.485 \text{ dB}$ (L9/2023, S75); However $K = 1.031 > 1$ so maximum gain is $MAG = |S_{21}| / |S_{12}| \cdot [K - \sqrt{K^2 - 1}] = 17.415 = 12.409 \text{ dB}$ (L9/2023, S76)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.087$, $\arg(S_{22}^*) = -150.4^\circ$; $\theta_{S1} = 122.7^\circ$; $\text{Im}(y_S) = -0.175$; **or**
 $\theta_{S2} = 27.7^\circ$; $\text{Im}(y_S) = 0.175$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.598$, $\arg(S_{11}^*) = 153.0^\circ$; $\theta_{S3} = 166.9^\circ$; $\text{Im}(y_S) = -1.492$; **or**
 $\theta_{S4} = 40.1^\circ$; $\text{Im}(y_S) = 1.492$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.175) + (-1.492) = -1.667$; $\theta_{S1} = 122.7^\circ$; $\theta_{P1} = 121.0^\circ$; $\theta_{S3} = 166.9^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.175) + (1.492) = 1.318$; $\theta_{S1} = 122.7^\circ$; $\theta_{P2} = 52.8^\circ$; $\theta_{S4} = 40.1^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.175) + (-1.492) = -1.318$; $\theta_{S2} = 27.7^\circ$; $\theta_{P3} = 127.2^\circ$; $\theta_{S3} = 166.9^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.175) + (1.492) = 1.667$; $\theta_{S2} = 27.7^\circ$; $\theta_{P4} = 59.0^\circ$; $\theta_{S4} = 40.1^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

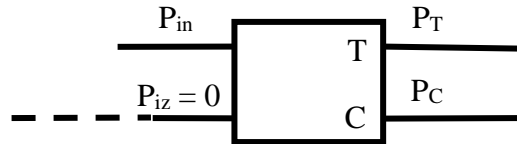
Subject no. 12

1. $z = 1.020 - j \cdot 0.720$; $Y = 1 / 50\Omega / (1.020 - j \cdot 0.720) = 0.0131S + j \cdot (0.0092)S$; $\Gamma = (z-1)/(z+1) = (1.020 - j \cdot 0.720 - 1)/(1.020 - j \cdot 0.720 + 1) = 0.122 + j \cdot (-0.313) = 0.336 \angle -68.8^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.130$, $Z_{0E} = 57.002\Omega$, $Z_{0O} = 43.858\Omega$

b) $P_c = 146.0\mu W = -8.356\text{dBm}$; $P_{in} = P_c + C = -8.356\text{dBm} + 17.7\text{dB} = 9.344\text{dBm} = 8.597 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [\text{mW}] = P_{in} - P_c - P_{iz} = 8.597\text{mW} - 0.1460\text{mW} - 0 = 8.451 \text{ mW} = 9.269 \text{ dBm}$



3. The shunt RC load with $R = 69\Omega$ and $C = 0.297\text{pF}$ has $Z_L = 30.60\Omega + j \cdot (-34.28)\Omega$ at 8.7GHz

$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$ (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 4/5) = -3.078$; $\cot(\beta l) = -0.325$; $Z_{in} = 328.92\Omega \angle 11.7^\circ = 322.07\Omega + j \cdot (66.77)\Omega$;

b) If the line becomes short-circuited $Z_L = 0$; $Z_{in} = j \cdot (-292.38)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 13\text{dB} + 11\text{dB} + 10\text{dB} = 34\text{dB}$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$F_1 = 2.66\text{dB} = 1.845$, $G_1 = 13\text{dB} = 19.953$, $F_2 = 2.54\text{dB} = 1.795$, $G_2 = 11\text{dB} = 12.589$, $F_3 = 2.41\text{dB} = 1.742$,
 $F = 1.845 + (1.795 - 1)/19.953 + (1.742 - 1)/19.953/12.589 = 1.888 = 2.760\text{dB}$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ' |
|----|-----------------------------|------------|-------|--------|
| T1 | $-0.303 + j \cdot (0.362)$ | 0.472 | 0.865 | 0.886 |
| T2 | $-0.184 + j \cdot (-0.031)$ | 0.187 | 1.058 | 1.026 |

b) $\mu'(T1) < \mu'(T2)$ so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.038$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 22.055 = 13.435 \text{ dB}$ (L9/2023, S75); However $K = 1.058 > 1$ so maximum gain is $MAG = |S_{21}| / |S_{12}| \cdot [K - \sqrt{K^2 - 1}] = 15.719 = 11.964 \text{ dB}$ (L9/2023, S76)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.087$, $\arg(S_{22}^*) = -131.2^\circ$; $\theta_{S1} = 113.1^\circ$; $\text{Im}(y_S) = -0.175$; **or**
 $\theta_{S2} = 18.1^\circ$; $\text{Im}(y_S) = 0.175$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.592$, $\arg(S_{11}^*) = 157.0^\circ$; $\theta_{S3} = 164.6^\circ$; $\text{Im}(y_S) = -1.469$; **or**
 $\theta_{S4} = 38.4^\circ$; $\text{Im}(y_S) = 1.469$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.175) + (-1.469) = -1.644$; $\theta_{S1} = 113.1^\circ$; $\theta_{P1} = 121.3^\circ$; $\theta_{S3} = 164.6^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.175) + (1.469) = 1.294$; $\theta_{S1} = 113.1^\circ$; $\theta_{P2} = 52.3^\circ$; $\theta_{S4} = 38.4^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.175) + (-1.469) = -1.294$; $\theta_{S2} = 18.1^\circ$; $\theta_{P3} = 127.7^\circ$; $\theta_{S3} = 164.6^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.175) + (1.469) = 1.644$; $\theta_{S2} = 18.1^\circ$; $\theta_{P4} = 58.7^\circ$; $\theta_{S4} = 38.4^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

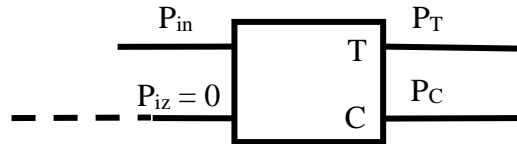
Subject no. 13

1. $z = 0.830 + j \cdot 1.195$; $Y = 1 / 50\Omega / (0.830 + j \cdot 1.195) = 0.0078S + j \cdot (-0.0113)S$; $\Gamma = (z-1)/(z+1) = (0.830 + j \cdot 1.195 - 1)/(0.830 + j \cdot 1.195 + 1) = 0.234 + j \cdot (0.500) = 0.552 \angle 65.0^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.115$, $Z_{0E} = 56.112\Omega$, $Z_{0O} = 44.554\Omega$

b) $P_c = 86.5\mu W = -10.630\text{dBm}$; $P_{in} = P_c + C = -10.630\text{dBm} + 18.8\text{dB} = 8.170\text{dBm} = 6.562 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [\text{mW}] = P_{in} - P_c - P_{iz} = 6.562\text{mW} - 0.0865\text{mW} - 0 = 6.475 \text{ mW} = 8.113 \text{ dBm}$



3. The shunt RL load with $R = 59\Omega$ and $L = 0.986\text{nH}$ has $Z_L = 19.14\Omega + j \cdot (27.62)\Omega$ at 6.6GHz

$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$ (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 2/3) = 1.732$; $\cot(\beta l) = 0.577$; $Z_{in} = 145.08\Omega \angle -15.2^\circ = 140.02\Omega + j \cdot (-37.99)\Omega$;

b) If the line becomes open-circuited $Z_L = \infty$; $Z_{in} = j \cdot (-25.98)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 15\text{dB} + 15\text{dB} + 14\text{dB} = 44\text{dB}$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$F_1 = 2.04\text{dB} = 1.600$, $G_1 = 15\text{dB} = 31.623$, $F_2 = 2.60\text{dB} = 1.820$, $G_2 = 15\text{dB} = 31.623$, $F_3 = 2.08\text{dB} = 1.614$,
 $F = 1.600 + (1.820 - 1)/31.623 + (1.614 - 1)/31.623/31.623 = 1.626 = 2.111\text{dB}$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ |
|----|-----------------------------|------------|-------|-------|
| T1 | $-0.336 + j \cdot (-0.365)$ | 0.496 | 0.553 | 0.431 |
| T2 | $-0.132 + j \cdot (-0.399)$ | 0.421 | 0.650 | 0.712 |

b) $\mu (T1) < \mu (T2)$ so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.350$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 32.484 = 15.117 \text{ dB}$ (L9/2023, S75)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.322$, $\arg(S_{22}^*) = 91.8^\circ$; $\theta_{S1} = 8.5^\circ$; $\text{Im}(y_S) = -0.680$; **or** $\theta_{S2} = 79.7^\circ$; $\text{Im}(y_S) = 0.680$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.763$, $\arg(S_{11}^*) = 88.0^\circ$; $\theta_{S3} = 25.9^\circ$; $\text{Im}(y_S) = -2.361$; **or** $\theta_{S4} = 66.1^\circ$; $\text{Im}(y_S) = 2.361$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.680) + (-2.361) = -3.041$; $\theta_{S1} = 8.5^\circ$; $\theta_{P1} = 108.2^\circ$; $\theta_{S3} = 25.9^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.680) + (2.361) = 1.681$; $\theta_{S1} = 8.5^\circ$; $\theta_{P2} = 59.2^\circ$; $\theta_{S4} = 66.1^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.680) + (-2.361) = -1.681$; $\theta_{S2} = 79.7^\circ$; $\theta_{P3} = 120.8^\circ$; $\theta_{S3} = 25.9^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.680) + (2.361) = 3.041$; $\theta_{S2} = 79.7^\circ$; $\theta_{P4} = 71.8^\circ$; $\theta_{S4} = 66.1^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

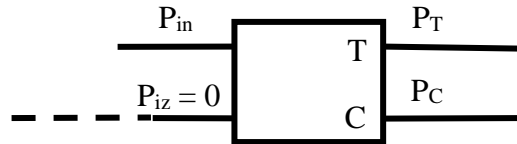
Subject no. 14

1. $z = 1.030 - j \cdot 1.085$; $Y = 1 / 50\Omega / (1.030 - j \cdot 1.085) = 0.0092S + j \cdot (0.0097)S$; $\Gamma = (z-1)/(z+1) = (1.030 - j \cdot 1.085 - 1)/(1.030 - j \cdot 1.085 + 1) = 0.234 + j \cdot (-0.410) = 0.472 \angle -60.3^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.080$, $Z_{0E} = 54.193\Omega$, $Z_{0O} = 46.132\Omega$

b) $P_c = 141.5\mu W = -8.492\text{dBm}$; $P_{in} = P_c + C = -8.492\text{dBm} + 21.9\text{dB} = 13.408\text{dBm} = 21.916 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [\text{mW}] = P_{in} - P_c - P_{iz} = 21.916\text{mW} - 0.1415\text{mW} - 0 = 21.774 \text{ mW} = 13.379 \text{ dBm}$



3. The shunt RC load with $R = 64\Omega$ and $C = 0.615\text{pF}$ has $Z_L = 11.59\Omega + j \cdot (-24.64)\Omega$ at 8.6GHz

$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$ (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 2/3) = 1.732$; $\cot(\beta l) = 0.577$; $Z_{in} = 16.49\Omega \angle 57.7^\circ = 8.82\Omega + j \cdot (13.93)\Omega$;

b) If the line becomes short-circuited $Z_L = 0$; $Z_{in} = j \cdot (60.62)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 17\text{dB} + 17\text{dB} + 16\text{dB} = 50\text{dB}$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$F_1 = 2.38\text{dB} = 1.730$, $G_1 = 17\text{dB} = 50.119$, $F_2 = 2.74\text{dB} = 1.879$, $G_2 = 17\text{dB} = 50.119$, $F_3 = 2.25\text{dB} = 1.679$,
 $F = 1.730 + (1.879 - 1)/50.119 + (1.679 - 1)/50.119/50.119 = 1.748 = 2.424\text{dB}$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ |
|----|-----------------------------|------------|-------|-------|
| T1 | $-0.318 + j \cdot (-0.380)$ | 0.496 | 0.547 | 0.426 |
| T2 | $-0.119 + j \cdot (-0.412)$ | 0.429 | 0.638 | 0.701 |

b) $\mu(T1) < \mu(T2)$ so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.366$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 33.000 = 15.185 \text{ dB}$ (L9/2023, S75)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.329$, $\arg(S_{22}^*) = 89.6^\circ$; $\theta_{S1} = 9.8^\circ$; $\text{Im}(y_S) = -0.697$; **or** $\theta_{S2} = 80.6^\circ$; $\text{Im}(y_S) = 0.697$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.770$, $\arg(S_{11}^*) = 86.0^\circ$; $\theta_{S3} = 27.2^\circ$; $\text{Im}(y_S) = -2.414$; **or** $\theta_{S4} = 66.8^\circ$; $\text{Im}(y_S) = 2.414$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.697) + (-2.414) = -3.110$; $\theta_{S1} = 9.8^\circ$; $\theta_{P1} = 107.8^\circ$; $\theta_{S3} = 27.2^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.697) + (2.414) = 1.717$; $\theta_{S1} = 9.8^\circ$; $\theta_{P2} = 59.8^\circ$; $\theta_{S4} = 66.8^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.697) + (-2.414) = -1.717$; $\theta_{S2} = 80.6^\circ$; $\theta_{P3} = 120.2^\circ$; $\theta_{S3} = 27.2^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.697) + (2.414) = 3.110$; $\theta_{S2} = 80.6^\circ$; $\theta_{P4} = 72.2^\circ$; $\theta_{S4} = 66.8^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

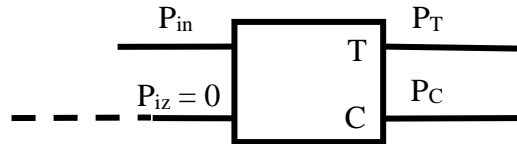
Subject no. 15

1. $z = 1.245 + j \cdot 1.150$; $Y = 1 / 50\Omega / (1.245 + j \cdot 1.150) = 0.0087S + j \cdot (-0.0080)S$; $\Gamma = (z-1)/(z+1) = (1.245 + j \cdot 1.150 - 1)/(1.245 + j \cdot 1.150 + 1) = 0.294 + j \cdot (0.361) = 0.466 \angle 50.8^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.088$, $Z_{0E} = 54.618\Omega$, $Z_{0O} = 45.773\Omega$

b) $P_c = 91.0\mu W = -10.410\text{dBm}$; $P_{in} = P_c + C = -10.410\text{dBm} + 21.1\text{dB} = 10.690\text{dBm} = 11.723 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [\text{mW}] = P_{in} - P_c - P_{iz} = 11.723\text{mW} - 0.0910\text{mW} - 0 = 11.632 \text{ mW} = 10.657 \text{ dBm}$



3. The shunt RC load with $R = 40\Omega$ and $C = 0.638\text{pF}$ has $Z_L = 17.42\Omega + j \cdot (-19.83)\Omega$ at 7.1GHz

$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$ (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 3/5) = 0.727$; $\cot(\beta l) = 1.376$; $Z_{in} = 16.04\Omega \angle 24.4^\circ = 14.61\Omega + j \cdot (6.63)\Omega$;

b) If the line becomes open-circuited $Z_L = \infty$; $Z_{in} = j \cdot (-61.94)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 16\text{dB} + 17\text{dB} + 18\text{dB} = 51\text{dB}$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$F_1 = 2.08\text{dB} = 1.614$, $G_1 = 16\text{dB} = 39.811$, $F_2 = 2.53\text{dB} = 1.791$, $G_2 = 17\text{dB} = 50.119$, $F_3 = 2.98\text{dB} = 1.986$,

$F = 1.614 + (1.791 - 1)/39.811 + (1.986 - 1)/39.811/50.119 = 1.635 = 2.134\text{dB}$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ |
|----|-----------------------------|------------|-------|-------|
| T1 | $-0.181 + j \cdot (0.450)$ | 0.484 | 0.908 | 0.845 |
| T2 | $-0.152 + j \cdot (-0.006)$ | 0.152 | 1.135 | 1.101 |

b) $\mu (T1) < \mu (T2)$ so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.044$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 21.255 = 13.275 \text{ dB}$ (L9/2023, S75); However $K = 1.135 > 1$ so maximum gain is $MAG = |S_{21}| / |S_{12}| \cdot [K - \sqrt{K^2 - 1}] = 12.703 = 11.039 \text{ dB}$ (L9/2023, S76)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.108$, $\arg(S_{22}^*) = -84.5^\circ$; $\theta_{S1} = 90.3^\circ$; $\text{Im}(y_S) = -0.217$; **or** $\theta_{S2} = 174.1^\circ$; $\text{Im}(y_S) = 0.217$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.578$, $\arg(S_{11}^*) = 169.0^\circ$; $\theta_{S3} = 158.2^\circ$; $\text{Im}(y_S) = -1.417$; **or** $\theta_{S4} = 32.8^\circ$; $\text{Im}(y_S) = 1.417$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.217) + (-1.417) = -1.634$; $\theta_{S1} = 90.3^\circ$; $\theta_{P1} = 121.5^\circ$; $\theta_{S3} = 158.2^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.217) + (1.417) = 1.199$; $\theta_{S1} = 90.3^\circ$; $\theta_{P2} = 50.2^\circ$; $\theta_{S4} = 32.8^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.217) + (-1.417) = -1.199$; $\theta_{S2} = 174.1^\circ$; $\theta_{P3} = 129.8^\circ$; $\theta_{S3} = 158.2^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.217) + (1.417) = 1.634$; $\theta_{S2} = 174.1^\circ$; $\theta_{P4} = 58.5^\circ$; $\theta_{S4} = 32.8^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

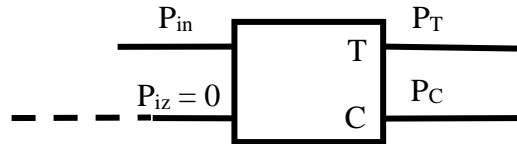
Subject no. 16

1. $z = 1.050 + j \cdot 0.890$; $Y = 1 / 50\Omega / (1.050 + j \cdot 0.890) = 0.0111S + j \cdot (-0.0094)S$; $\Gamma = (z-1)/(z+1) = (1.050 + j \cdot 0.890 - 1)/(1.050 + j \cdot 0.890 + 1) = 0.179 + j \cdot (0.356) = 0.399 \angle 63.3^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.094$, $Z_{0E} = 54.966\Omega$, $Z_{0O} = 45.483\Omega$

b) $P_c = 112.5\mu W = -9.488\text{dBm}$; $P_{in} = P_c + C = -9.488\text{dBm} + 20.5\text{dB} = 11.012\text{dBm} = 12.623 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [\text{mW}] = P_{in} - P_c - P_{iz} = 12.623\text{mW} - 0.1125\text{mW} - 0 = 12.510 \text{ mW} = 10.973 \text{ dBm}$



3. The shunt RL load with $R = 70\Omega$ and $L = 0.961\text{nH}$ has $Z_L = 21.04\Omega + j \cdot (32.10)\Omega$ at 7.6GHz

$$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$$
 (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 2/6) = -1.732$; $\cot(\beta l) = -0.577$; $Z_{in} = 16.71\Omega \angle -39.6^\circ = 12.87\Omega + j \cdot (-10.65)\Omega$;

b) If the line becomes open-circuited $Z_L = \infty$; $Z_{in} = j \cdot (23.09)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 15\text{dB} + 17\text{dB} + 11\text{dB} = 43\text{dB}$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$F_1 = 2.06\text{dB} = 1.607$, $G_1 = 15\text{dB} = 31.623$, $F_2 = 2.04\text{dB} = 1.600$, $G_2 = 17\text{dB} = 50.119$, $F_3 = 2.36\text{dB} = 1.722$,
 $F = 1.607 + (1.600 - 1)/31.623 + (1.722 - 1)/31.623/50.119 = 1.626 = 2.112\text{dB}$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ |
|----|-----------------------------|------------|-------|-------|
| T1 | $-0.463 + j \cdot (0.105)$ | 0.475 | 0.763 | 0.652 |
| T2 | $-0.233 + j \cdot (-0.150)$ | 0.277 | 0.887 | 0.912 |

b) $\mu (T1) < \mu (T2)$ so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.086$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 23.909 = 13.786 \text{ dB}$ (L9/2023, S75)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.151$, $\arg(S_{22}^*) = 143.1^\circ$; $\theta_{S1} = 157.8^\circ$; $\text{Im}(y_S) = -0.306$; **or**
 $\theta_{S2} = 59.1^\circ$; $\text{Im}(y_S) = 0.306$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.634$, $\arg(S_{11}^*) = 129.8^\circ$; $\theta_{S3} = 179.8^\circ$; $\text{Im}(y_S) = -1.640$; **or**
 $\theta_{S4} = 50.4^\circ$; $\text{Im}(y_S) = 1.640$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.306) + (-1.640) = -1.945$; $\theta_{S1} = 157.8^\circ$; $\theta_{P1} = 117.2^\circ$; $\theta_{S3} = 179.8^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.306) + (1.640) = 1.334$; $\theta_{S1} = 157.8^\circ$; $\theta_{P2} = 53.1^\circ$; $\theta_{S4} = 50.4^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.306) + (-1.640) = -1.334$; $\theta_{S2} = 59.1^\circ$; $\theta_{P3} = 126.9^\circ$; $\theta_{S3} = 179.8^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.306) + (1.640) = 1.945$; $\theta_{S2} = 59.1^\circ$; $\theta_{P4} = 62.8^\circ$; $\theta_{S4} = 50.4^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

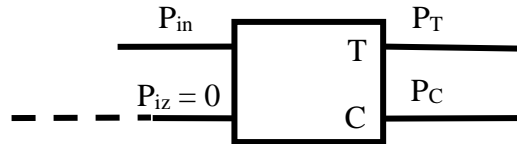
Subject no. 17

1. $z = 0.750 + j \cdot 0.940$; $Y = 1 / 50\Omega / (0.750 + j \cdot 0.940) = 0.0104S + j \cdot (-0.0130)S$; $\Gamma = (z-1)/(z+1) = (0.750 + j \cdot 0.940 - 1)/(0.750 + j \cdot 0.940 + 1) = 0.113 + j \cdot (0.476) = 0.490 \angle 76.7^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.105$, $Z_{0E} = 55.541\Omega$, $Z_{0O} = 45.012\Omega$

b) $P_c = 126.5\mu W = -8.979\text{dBm}$; $P_{in} = P_c + C = -8.979\text{dBm} + 19.6\text{dB} = 10.621\text{dBm} = 11.537 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [\text{mW}] = P_{in} - P_c - P_{iz} = 11.537\text{mW} - 0.1265\text{mW} - 0 = 11.410 \text{ mW} = 10.573 \text{ dBm}$



3. The series RC load with $R = 70\Omega$ and $C = 0.609\text{pF}$ has $Z_L = 70.00\Omega + j \cdot (-38.43)\Omega$ at 6.8GHz

$$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$$
 (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 2/8) = \infty$; $\cot(\beta l) = 0.000$; $Z_{in} = 15.34\Omega \angle 28.8^\circ = 13.45\Omega + j \cdot (7.38)\Omega$;

b) If the line becomes short-circuited $Z_L = 0$; $Z_{in} = j \cdot (0.00)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 16\text{dB} + 19\text{dB} + 19\text{dB} = 54\text{dB}$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$F_1 = 2.65\text{dB} = 1.841$, $G_1 = 16\text{dB} = 39.811$, $F_2 = 2.97\text{dB} = 1.982$, $G_2 = 19\text{dB} = 79.433$, $F_3 = 2.70\text{dB} = 1.862$,
 $F = 1.841 + (1.982 - 1)/39.811 + (1.862 - 1)/39.811/79.433 = 1.866 = 2.708\text{dB}$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ |
|----|-----------------------------|------------|-------|-------|
| T1 | $-0.464 + j \cdot (-0.155)$ | 0.489 | 0.633 | 0.507 |
| T2 | $-0.213 + j \cdot (-0.272)$ | 0.346 | 0.773 | 0.819 |

b) $\mu (T1) < \mu (T2)$ so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.209$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 27.851 = 14.448 \text{ dB}$ (L9/2023, S75)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.251$, $\arg(S_{22}^*) = 114.2^\circ$; $\theta_{S1} = 175.2^\circ$; $\text{Im}(y_S) = -0.519$; **or**
 $\theta_{S2} = 70.6^\circ$; $\text{Im}(y_S) = 0.519$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.694$, $\arg(S_{11}^*) = 108.0^\circ$; $\theta_{S3} = 13.0^\circ$; $\text{Im}(y_S) = -1.928$; **or**
 $\theta_{S4} = 59.0^\circ$; $\text{Im}(y_S) = 1.928$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.519) + (-1.928) = -2.446$; $\theta_{S1} = 175.2^\circ$; $\theta_{P1} = 112.2^\circ$; $\theta_{S3} = 13.0^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.519) + (1.928) = 1.409$; $\theta_{S1} = 175.2^\circ$; $\theta_{P2} = 54.6^\circ$; $\theta_{S4} = 59.0^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.519) + (-1.928) = -1.409$; $\theta_{S2} = 70.6^\circ$; $\theta_{P3} = 125.4^\circ$; $\theta_{S3} = 13.0^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.519) + (1.928) = 2.446$; $\theta_{S2} = 70.6^\circ$; $\theta_{P4} = 67.8^\circ$; $\theta_{S4} = 59.0^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

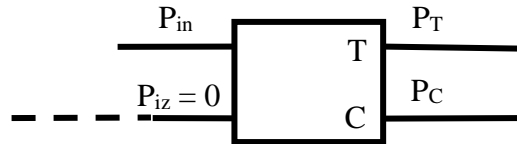
Subject no. 18

1. $z = 1.170 + j \cdot 0.870$; $Y = 1 / 50\Omega / (1.170 + j \cdot 0.870) = 0.0110S + j \cdot (-0.0082)S$; $\Gamma = (z-1)/(z+1) = (1.170 + j \cdot 0.870 - 1)/(1.170 + j \cdot 0.870 + 1) = 0.206 + j \cdot (0.318) = 0.379 \angle 57.1^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.091$, $Z_{0E} = 54.788\Omega$, $Z_{0O} = 45.630\Omega$

b) $P_c = 147.0\mu W = -8.327\text{dBm}$; $P_{in} = P_c + C = -8.327\text{dBm} + 20.8\text{dB} = 12.473\text{dBm} = 17.673 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [\text{mW}] = P_{in} - P_c - P_{iz} = 17.673\text{mW} - 0.1470\text{mW} - 0 = 17.526 \text{ mW} = 12.437 \text{ dBm}$



3. The shunt RL load with $R = 29\Omega$ and $L = 1.240\text{nH}$ has $Z_L = 24.84\Omega + j \cdot (10.16)\Omega$ at 9.1GHz

$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$ (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 2/5) = -0.727$; $\cot(\beta l) = -1.376$; $Z_{in} = 27.24\Omega \angle -23.2^\circ = 25.04\Omega + j \cdot (-10.73)\Omega$;

b) If the line becomes open-circuited $Z_L = \infty$; $Z_{in} = j \cdot (61.94)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 16\text{dB} + 12\text{dB} + 12\text{dB} = 40\text{dB}$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$F_1 = 2.49\text{dB} = 1.774$, $G_1 = 16\text{dB} = 39.811$, $F_2 = 2.12\text{dB} = 1.629$, $G_2 = 12\text{dB} = 15.849$, $F_3 = 2.46\text{dB} = 1.762$,
 $F = 1.774 + (1.629 - 1)/39.811 + (1.762 - 1)/39.811/15.849 = 1.791 = 2.531\text{dB}$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ' |
|----|-----------------------------|------------|-------|--------|
| T1 | $-0.422 + j \cdot (0.224)$ | 0.478 | 0.820 | 0.853 |
| T2 | $-0.220 + j \cdot (-0.089)$ | 0.237 | 0.955 | 0.979 |

b) $\mu'(T1) < \mu'(T2)$ so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.050$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 23.091 = 13.634 \text{ dB}$ (L9/2023, S75)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.102$, $\arg(S_{22}^*) = 164.3^\circ$; $\theta_{S1} = 145.8^\circ$; $\text{Im}(y_S) = -0.205$; **or**
 $\theta_{S2} = 49.9^\circ$; $\text{Im}(y_S) = 0.205$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.616$, $\arg(S_{11}^*) = 141.2^\circ$; $\theta_{S3} = 173.4^\circ$; $\text{Im}(y_S) = -1.564$; **or**
 $\theta_{S4} = 45.4^\circ$; $\text{Im}(y_S) = 1.564$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.205) + (-1.564) = -1.769$; $\theta_{S1} = 145.8^\circ$; $\theta_{P1} = 119.5^\circ$; $\theta_{S3} = 173.4^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.205) + (1.564) = 1.359$; $\theta_{S1} = 145.8^\circ$; $\theta_{P2} = 53.7^\circ$; $\theta_{S4} = 45.4^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.205) + (-1.564) = -1.359$; $\theta_{S2} = 49.9^\circ$; $\theta_{P3} = 126.3^\circ$; $\theta_{S3} = 173.4^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.205) + (1.564) = 1.769$; $\theta_{S2} = 49.9^\circ$; $\theta_{P4} = 60.5^\circ$; $\theta_{S4} = 45.4^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

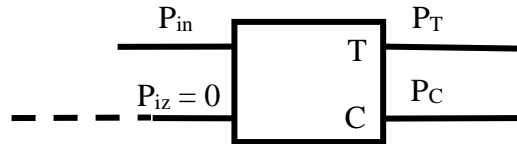
Subject no. 19

1. $z = 0.795 + j \cdot 1.145$; $Y = 1 / 50\Omega / (0.795 + j \cdot 1.145) = 0.0082S + j \cdot (-0.0118)S$; $\Gamma = (z-1)/(z+1) = (0.795 + j \cdot 1.145 - 1)/(0.795 + j \cdot 1.145 + 1) = 0.208 + j \cdot (0.505) = 0.546 \angle 67.6^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.094$, $Z_{0E} = 54.966\Omega$, $Z_{0O} = 45.483\Omega$

b) $P_c = 131.5\mu W = -8.811\text{dBm}$; $P_{in} = P_c + C = -8.811\text{dBm} + 20.5\text{dB} = 11.689\text{dBm} = 14.755 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [\text{mW}] = P_{in} - P_c - P_{iz} = 14.755\text{mW} - 0.1315\text{mW} - 0 = 14.623 \text{ mW} = 11.650 \text{ dBm}$



3. The series RC load with $R = 30\Omega$ and $C = 0.454\text{pF}$ has $Z_L = 30.00\Omega + j \cdot (-45.53)\Omega$ at 7.7GHz

$$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$$
 (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 7/8) = -1.000$; $\cot(\beta l) = -1.000$; $Z_{in} = 215.81\Omega \angle -24.6^\circ = 196.14\Omega + j \cdot (-90.00)\Omega$;

b) If the line becomes short-circuited $Z_L = 0$; $Z_{in} = j \cdot (-70.00)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 11\text{dB} + 18\text{dB} + 10\text{dB} = 39\text{dB}$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$F_1 = 2.04\text{dB} = 1.600$, $G_1 = 11\text{dB} = 12.589$, $F_2 = 2.58\text{dB} = 1.811$, $G_2 = 18\text{dB} = 63.096$, $F_3 = 2.15\text{dB} = 1.641$,
 $F = 1.600 + (1.811 - 1)/12.589 + (1.641 - 1)/12.589/63.096 = 1.665 = 2.214\text{dB}$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ' |
|----|-----------------------------|------------|-------|--------|
| T1 | $-0.480 + j \cdot (0.016)$ | 0.480 | 0.720 | 0.794 |
| T2 | $-0.232 + j \cdot (-0.193)$ | 0.302 | 0.845 | 0.926 |

b) $\mu'(T1) < \mu'(T2)$ so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.120$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 24.944 = 13.970 \text{ dB}$ (L9/2023, S75)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.186$, $\arg(S_{22}^*) = 131.2^\circ$; $\theta_{S1} = 164.7^\circ$; $\text{Im}(y_S) = -0.379$; **or** $\theta_{S2} = 64.0^\circ$; $\text{Im}(y_S) = 0.379$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.652$, $\arg(S_{11}^*) = 122.0^\circ$; $\theta_{S3} = 4.3^\circ$; $\text{Im}(y_S) = -1.720$; **or** $\theta_{S4} = 53.7^\circ$; $\text{Im}(y_S) = 1.720$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.379) + (-1.720) = -2.098$; $\theta_{S1} = 164.7^\circ$; $\theta_{P1} = 115.5^\circ$; $\theta_{S3} = 4.3^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.379) + (1.720) = 1.341$; $\theta_{S1} = 164.7^\circ$; $\theta_{P2} = 53.3^\circ$; $\theta_{S4} = 53.7^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.379) + (-1.720) = -1.341$; $\theta_{S2} = 64.0^\circ$; $\theta_{P3} = 126.7^\circ$; $\theta_{S3} = 4.3^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.379) + (1.720) = 2.098$; $\theta_{S2} = 64.0^\circ$; $\theta_{P4} = 64.5^\circ$; $\theta_{S4} = 53.7^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

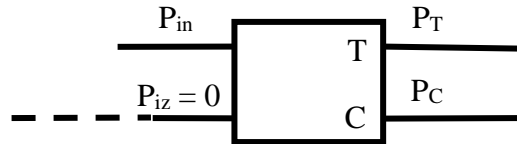
Subject no. 20

1. $z = 1.070 + j \cdot 1.035$; $Y = 1 / 50\Omega / (1.070 + j \cdot 1.035) = 0.0097S + j \cdot (-0.0093)S$; $\Gamma = (z-1)/(z+1) = (1.070 + j \cdot 1.035 - 1)/(1.070 + j \cdot 1.035 + 1) = 0.227 + j \cdot (0.386) = 0.448 \angle 59.6^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.107$, $Z_{0E} = 55.678\Omega$, $Z_{0O} = 44.901\Omega$

b) $P_c = 140.5\mu W = -8.523\text{dBm}$; $P_{in} = P_c + C = -8.523\text{dBm} + 19.4\text{dB} = 10.877\text{dBm} = 12.237 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [\text{mW}] = P_{in} - P_c - P_{iz} = 12.237\text{mW} - 0.1405\text{mW} - 0 = 12.097 \text{ mW} = 10.827 \text{ dBm}$



3. The shunt RC load with $R = 57\Omega$ and $C = 0.281\text{pF}$ has $Z_L = 32.91\Omega + j \cdot (-28.16)\Omega$ at 8.5GHz

$$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$$
 (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 2/6) = -1.732$; $\cot(\beta l) = -0.577$; $Z_{in} = 158.49\Omega \angle -2.7^\circ = 158.31\Omega + j \cdot (-7.55)\Omega$;

b) If the line becomes short-circuited $Z_L = 0$; $Z_{in} = j \cdot (-112.58)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 14\text{dB} + 14\text{dB} + 14\text{dB} = 42\text{dB}$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$F_1 = 2.28\text{dB} = 1.690$, $G_1 = 14\text{dB} = 25.119$, $F_2 = 2.82\text{dB} = 1.914$, $G_2 = 14\text{dB} = 25.119$, $F_3 = 2.26\text{dB} = 1.683$,
 $F = 1.690 + (1.914 - 1)/25.119 + (1.683 - 1)/25.119/25.119 = 1.728 = 2.375\text{dB}$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ' |
|----|-----------------------------|------------|-------|--------|
| T1 | $-0.385 + j \cdot (0.282)$ | 0.477 | 0.843 | 0.869 |
| T2 | $-0.208 + j \cdot (-0.064)$ | 0.218 | 0.993 | 0.997 |

b) $\mu' (T1) < \mu' (T2)$ so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.040$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 22.691 = 13.559 \text{ dB}$ (L9/2023, S75)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.086$, $\arg(S_{22}^*) = -179.1^\circ$; $\theta_{S1} = 137.0^\circ$; $\text{Im}(y_S) = -0.173$; **or**
 $\theta_{S2} = 42.1^\circ$; $\text{Im}(y_S) = 0.173$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.607$, $\arg(S_{11}^*) = 147.0^\circ$; $\theta_{S3} = 170.2^\circ$; $\text{Im}(y_S) = -1.528$; **or**
 $\theta_{S4} = 42.8^\circ$; $\text{Im}(y_S) = 1.528$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.173) + (-1.528) = -1.700$; $\theta_{S1} = 137.0^\circ$; $\theta_{P1} = 120.5^\circ$; $\theta_{S3} = 170.2^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.173) + (1.528) = 1.355$; $\theta_{S1} = 137.0^\circ$; $\theta_{P2} = 53.6^\circ$; $\theta_{S4} = 42.8^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.173) + (-1.528) = -1.355$; $\theta_{S2} = 42.1^\circ$; $\theta_{P3} = 126.4^\circ$; $\theta_{S3} = 170.2^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.173) + (1.528) = 1.700$; $\theta_{S2} = 42.1^\circ$; $\theta_{P4} = 59.5^\circ$; $\theta_{S4} = 42.8^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

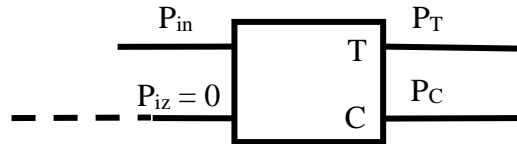
Subject no. 21

1. $z = 0.935 + j \cdot 1.265$; $Y = 1 / 50\Omega / (0.935 + j \cdot 1.265) = 0.0076S + j \cdot (-0.0102)S$; $\Gamma = (z-1)/(z+1) = (0.935 + j \cdot 1.265 - 1)/(0.935 + j \cdot 1.265 + 1) = 0.276 + j \cdot (0.473) = 0.548 \angle 59.8^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.079$, $Z_{0E} = 54.143\Omega$, $Z_{0O} = 46.174\Omega$

b) $P_c = 80.0\mu W = -10.969\text{dBm}$; $P_{in} = P_c + C = -10.969\text{dBm} + 22.0\text{dB} = 11.031\text{dBm} = 12.679 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [\text{mW}] = P_{in} - P_c - P_{iz} = 12.679\text{mW} - 0.0800\text{mW} - 0 = 12.599 \text{ mW} = 11.003 \text{ dBm}$



3. The shunt RC load with $R = 62\Omega$ and $C = 0.304\text{pF}$ has $Z_L = 28.02\Omega + j \cdot (-30.86)\Omega$ at 9.3GHz

$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$ (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 6/8) = \infty$; $\cot(\beta l) = 0.000$; $Z_{in} = 59.98\Omega \angle 47.8^\circ = 40.32\Omega + j \cdot (44.41)\Omega$;

b) If the line becomes open-circuited $Z_L = \infty$; $Z_{in} = j \cdot (0.00)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 11\text{dB} + 13\text{dB} + 11\text{dB} = 35\text{dB}$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$F_1 = 2.26\text{dB} = 1.683$, $G_1 = 11\text{dB} = 12.589$, $F_2 = 2.41\text{dB} = 1.742$, $G_2 = 13\text{dB} = 19.953$, $F_3 = 2.30\text{dB} = 1.698$,
 $F = 1.683 + (1.742 - 1)/12.589 + (1.698 - 1)/12.589/19.953 = 1.744 = 2.416\text{dB}$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ' |
|----|-----------------------------|------------|-------|--------|
| T1 | $-0.352 + j \cdot (-0.344)$ | 0.493 | 0.561 | 0.736 |
| T2 | $-0.144 + j \cdot (-0.387)$ | 0.413 | 0.662 | 0.858 |

b) μ' (T1) < μ' (T2) so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.332$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 31.978 = 15.049 \text{ dB}$ (L9/2023, S75)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.315$, $\arg(S_{22}^*) = 94.1^\circ$; $\theta_{S1} = 7.2^\circ$; $\text{Im}(y_S) = -0.664$; **or** $\theta_{S2} = 78.8^\circ$; $\text{Im}(y_S) = 0.664$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.756$, $\arg(S_{11}^*) = 90.0^\circ$; $\theta_{S3} = 24.6^\circ$; $\text{Im}(y_S) = -2.310$; **or** $\theta_{S4} = 65.4^\circ$; $\text{Im}(y_S) = 2.310$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.664) + (-2.310) = -2.974$; $\theta_{S1} = 7.2^\circ$; $\theta_{P1} = 108.6^\circ$; $\theta_{S3} = 24.6^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.664) + (2.310) = 1.646$; $\theta_{S1} = 7.2^\circ$; $\theta_{P2} = 58.7^\circ$; $\theta_{S4} = 65.4^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.664) + (-2.310) = -1.646$; $\theta_{S2} = 78.8^\circ$; $\theta_{P3} = 121.3^\circ$; $\theta_{S3} = 24.6^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.664) + (2.310) = 2.974$; $\theta_{S2} = 78.8^\circ$; $\theta_{P4} = 71.4^\circ$; $\theta_{S4} = 65.4^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

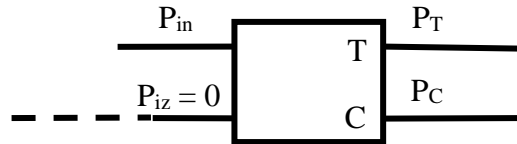
Subject no. 22

1. $z = 1.280 - j \cdot 1.020$; $Y = 1 / 50\Omega / (1.280 - j \cdot 1.020) = 0.0096S + j \cdot (0.0076)S$; $\Gamma = (z-1)/(z+1) = (1.280 - j \cdot 1.020 - 1)/(1.280 - j \cdot 1.020 + 1) = 0.269 + j \cdot (-0.327) = 0.423 \angle -50.5^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.124$, $Z_{0E} = 56.663\Omega$, $Z_{0O} = 44.120\Omega$

b) $P_c = 75.5\mu W = -11.221\text{dBm}$; $P_{in} = P_c + C = -11.221\text{dBm} + 18.1\text{dB} = 6.879\text{dBm} = 4.875 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [\text{mW}] = P_{in} - P_c - P_{iz} = 4.875\text{mW} - 0.0755\text{mW} - 0 = 4.799 \text{ mW} = 6.812 \text{ dBm}$



3. The shunt RL load with $R = 38\Omega$ and $L = 1.202\text{nH}$ has $Z_L = 29.94\Omega + j \cdot (15.53)\Omega$ at 9.7GHz

$$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$$
 (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 4/6) = 1.732$; $\cot(\beta l) = 0.577$; $Z_{in} = 202.60\Omega \angle 43.3^\circ = 147.53\Omega + j \cdot (138.86)\Omega$;

b) If the line becomes open-circuited $Z_L = \infty$; $Z_{in} = j \cdot (-54.85)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 11\text{dB} + 17\text{dB} + 14\text{dB} = 42\text{dB}$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$$F_1 = 2.79\text{dB} = 1.901, G_1 = 11\text{dB} = 12.589, F_2 = 2.72\text{dB} = 1.871, G_2 = 17\text{dB} = 50.119, F_3 = 2.71\text{dB} = 1.866, F = 1.901 + (1.871 - 1)/12.589 + (1.866 - 1)/12.589/50.119 = 1.972 = 2.948\text{dB}$$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ |
|----|-----------------------------|------------|-------|-------|
| T1 | $-0.401 + j \cdot (0.265)$ | 0.481 | 0.844 | 0.750 |
| T2 | $-0.213 + j \cdot (-0.072)$ | 0.225 | 0.981 | 0.985 |

b) $\mu(T1) < \mu(T2)$ so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.040$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 22.818 = 13.583 \text{ dB}$ (L9/2023, S75)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.086$, $\arg(S_{22}^*) = 171.3^\circ$; $\theta_{S1} = 141.8^\circ$; $\text{Im}(y_S) = -0.173$; **or** $\theta_{S2} = 46.9^\circ$; $\text{Im}(y_S) = 0.173$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.610$, $\arg(S_{11}^*) = 145.0^\circ$; $\theta_{S3} = 171.3^\circ$; $\text{Im}(y_S) = -1.540$; **or** $\theta_{S4} = 43.7^\circ$; $\text{Im}(y_S) = 1.540$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.173) + (-1.540) = -1.712$; $\theta_{S1} = 141.8^\circ$; $\theta_{P1} = 120.3^\circ$; $\theta_{S3} = 171.3^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.173) + (1.540) = 1.367$; $\theta_{S1} = 141.8^\circ$; $\theta_{P2} = 53.8^\circ$; $\theta_{S4} = 43.7^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.173) + (-1.540) = -1.367$; $\theta_{S2} = 46.9^\circ$; $\theta_{P3} = 126.2^\circ$; $\theta_{S3} = 171.3^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.173) + (1.540) = 1.712$; $\theta_{S2} = 46.9^\circ$; $\theta_{P4} = 59.7^\circ$; $\theta_{S4} = 43.7^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

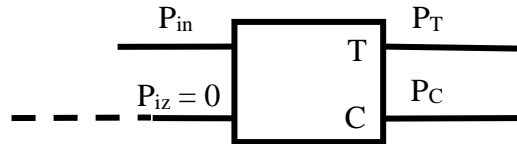
Subject no. 23

1. $z = 0.725 - j \cdot 1.165$; $Y = 1 / 50\Omega / (0.725 - j \cdot 1.165) = 0.0077S + j \cdot (0.0124)S$; $\Gamma = (z-1)/(z+1) = (0.725 - j \cdot 1.165 - 1)/(0.725 - j \cdot 1.165 + 1) = 0.204 + j \cdot (-0.538) = 0.575 \angle -69.2^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.075$, $Z_{0E} = 53.901\Omega$, $Z_{0O} = 46.381\Omega$

b) $P_c = 68.5\mu W = -11.643\text{dBm}$; $P_{in} = P_c + C = -11.643\text{dBm} + 22.5\text{dB} = 10.857\text{dBm} = 12.181 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [\text{mW}] = P_{in} - P_c - P_{iz} = 12.181\text{mW} - 0.0685\text{mW} - 0 = 12.113 \text{ mW} = 10.832 \text{ dBm}$



3. The series RL load with $R = 47\Omega$ and $L = 1.369\text{nH}$ has $Z_L = 47.00\Omega + j \cdot (67.95)\Omega$ at 7.9GHz

$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$ (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 2/5) = -0.727$; $\cot(\beta l) = -1.376$; $Z_{in} = 26.96\Omega \angle 52.9^\circ = 16.26\Omega + j \cdot (21.50)\Omega$;

b) If the line becomes open-circuited $Z_L = \infty$; $Z_{in} = j \cdot (68.82)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 17\text{dB} + 12\text{dB} + 11\text{dB} = 40\text{dB}$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$F_1 = 2.76\text{dB} = 1.888$, $G_1 = 17\text{dB} = 50.119$, $F_2 = 2.49\text{dB} = 1.774$, $G_2 = 12\text{dB} = 15.849$, $F_3 = 2.87\text{dB} = 1.936$,
 $F = 1.888 + (1.774 - 1)/50.119 + (1.936 - 1)/50.119/15.849 = 1.905 = 2.798\text{dB}$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ' |
|----|-----------------------------|------------|-------|--------|
| T1 | $-0.204 + j \cdot (0.442)$ | 0.487 | 0.905 | 0.916 |
| T2 | $-0.158 + j \cdot (-0.010)$ | 0.158 | 1.125 | 1.056 |

b) $\mu'(T1) < \mu'(T2)$ so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.040$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 21.400 = 13.304 \text{ dB}$ (L9/2023, S75); However $K = 1.125 > 1$ so maximum gain is $MAG = |S_{21}| / |S_{12}| \cdot [K - \sqrt{K^2 - 1}] = 13.035 = 11.151 \text{ dB}$ (L9/2023, S76)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.098$, $\arg(S_{22}^*) = -88.7^\circ$; $\theta_{S1} = 92.2^\circ$; $\text{Im}(y_S) = -0.197$; **or**
 $\theta_{S2} = 176.5^\circ$; $\text{Im}(y_S) = 0.197$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.579$, $\arg(S_{11}^*) = 167.0^\circ$; $\theta_{S3} = 159.2^\circ$; $\text{Im}(y_S) = -1.420$; **or**
 $\theta_{S4} = 33.8^\circ$; $\text{Im}(y_S) = 1.420$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.197) + (-1.420) = -1.617$; $\theta_{S1} = 92.2^\circ$; $\theta_{P1} = 121.7^\circ$; $\theta_{S3} = 159.2^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.197) + (1.420) = 1.223$; $\theta_{S1} = 92.2^\circ$; $\theta_{P2} = 50.7^\circ$; $\theta_{S4} = 33.8^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.197) + (-1.420) = -1.223$; $\theta_{S2} = 176.5^\circ$; $\theta_{P3} = 129.3^\circ$; $\theta_{S3} = 159.2^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.197) + (1.420) = 1.617$; $\theta_{S2} = 176.5^\circ$; $\theta_{P4} = 58.3^\circ$; $\theta_{S4} = 33.8^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

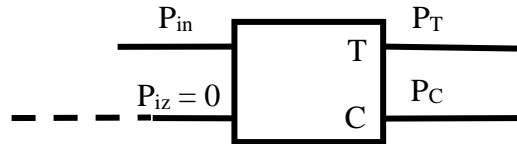
Subject no. 24

1. $z = 0.865 + j \cdot 1.175$; $Y = 1 / 50\Omega / (0.865 + j \cdot 1.175) = 0.0081S + j \cdot (-0.0110)S$; $\Gamma = (z-1)/(z+1) = (0.865 + j \cdot 1.175 - 1)/(0.865 + j \cdot 1.175 + 1) = 0.232 + j \cdot (0.484) = 0.537 \angle 64.3^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.107$, $Z_{0E} = 55.678\Omega$, $Z_{0O} = 44.901\Omega$

b) $P_c = 90.0\mu W = -10.458\text{dBm}$; $P_{in} = P_c + C = -10.458\text{dBm} + 19.4\text{dB} = 8.942\text{dBm} = 7.839 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [\text{mW}] = P_{in} - P_c - P_{iz} = 7.839\text{mW} - 0.0900\text{mW} - 0 = 7.749 \text{ mW} = 8.892 \text{ dBm}$



3. The shunt RC load with $R = 68\Omega$ and $C = 0.320\text{pF}$ has $Z_L = 24.33\Omega + j \cdot (-32.59)\Omega$ at 9.8GHz

$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$ (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 4/8) = 0.000$; $\cot(\beta l) = \infty$; $Z_{in} = 40.67\Omega \angle -53.3^\circ = 24.33\Omega + j \cdot (-32.59)\Omega$;

b) If the line becomes short-circuited $Z_L = 0$; $Z_{in} = j \cdot (0.00)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 10\text{dB} + 14\text{dB} + 13\text{dB} = 37\text{dB}$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$F_1 = 2.71\text{dB} = 1.866$, $G_1 = 10\text{dB} = 10.000$, $F_2 = 2.64\text{dB} = 1.837$, $G_2 = 14\text{dB} = 25.119$, $F_3 = 2.88\text{dB} = 1.941$,
 $F = 1.866 + (1.837 - 1)/10.000 + (1.941 - 1)/10.000/25.119 = 1.954 = 2.909\text{dB}$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ' |
|----|-----------------------------|------------|-------|--------|
| T1 | $-0.479 + j \cdot (0.041)$ | 0.480 | 0.733 | 0.801 |
| T2 | $-0.233 + j \cdot (-0.183)$ | 0.297 | 0.856 | 0.931 |

b) $\mu' (T1) < \mu' (T2)$ so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.110$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 24.560 = 13.902 \text{ dB}$ (L9/2023, S75)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.176$, $\arg(S_{22}^*) = 133.7^\circ$; $\theta_{S1} = 163.2^\circ$; $\text{Im}(y_S) = -0.358$; **or**
 $\theta_{S2} = 63.1^\circ$; $\text{Im}(y_S) = 0.358$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.646$, $\arg(S_{11}^*) = 124.0^\circ$; $\theta_{S3} = 3.1^\circ$; $\text{Im}(y_S) = -1.693$; **or** $\theta_{S4} = 52.9^\circ$; $\text{Im}(y_S) = 1.693$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.358) + (-1.693) = -2.050$; $\theta_{S1} = 163.2^\circ$; $\theta_{P1} = 116.0^\circ$; $\theta_{S3} = 3.1^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.358) + (1.693) = 1.335$; $\theta_{S1} = 163.2^\circ$; $\theta_{P2} = 53.2^\circ$; $\theta_{S4} = 52.9^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.358) + (-1.693) = -1.335$; $\theta_{S2} = 63.1^\circ$; $\theta_{P3} = 126.8^\circ$; $\theta_{S3} = 3.1^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.358) + (1.693) = 2.050$; $\theta_{S2} = 63.1^\circ$; $\theta_{P4} = 64.0^\circ$; $\theta_{S4} = 52.9^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

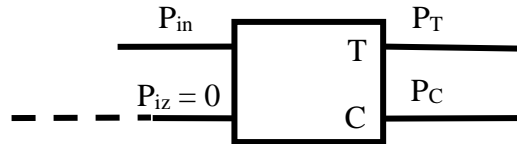
Subject no. 25

1. $z = 0.875 + j \cdot 0.990$; $Y = 1 / 50\Omega / (0.875 + j \cdot 0.990) = 0.0100S + j \cdot (-0.0113)S$; $\Gamma = (z-1)/(z+1) = (0.875 + j \cdot 0.990 - 1)/(0.875 + j \cdot 0.990 + 1) = 0.166 + j \cdot (0.440) = 0.471 \angle 69.4^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.105$, $Z_{0E} = 55.541\Omega$, $Z_{0O} = 45.012\Omega$

b) $P_c = 79.5\mu W = -10.996\text{dBm}$; $P_{in} = P_c + C = -10.996\text{dBm} + 19.6\text{dB} = 8.604\text{dBm} = 7.250 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [\text{mW}] = P_{in} - P_c - P_{iz} = 7.250\text{mW} - 0.0795\text{mW} - 0 = 7.171 \text{ mW} = 8.556 \text{ dBm}$



3. The series RC load with $R = 39\Omega$ and $C = 0.598\text{pF}$ has $Z_L = 39.00\Omega + j \cdot (-36.46)\Omega$ at 7.3GHz

$$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$$
 (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 3/6) = 0.000$; $\cot(\beta l) = \infty$; $Z_{in} = 53.39\Omega \angle -43.1^\circ = 39.00\Omega + j \cdot (-36.46)\Omega$;

b) If the line becomes open-circuited $Z_L = \infty$; $Z_{in} = j \cdot (0.00)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 14\text{dB} + 16\text{dB} + 15\text{dB} = 45\text{dB}$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$F_1 = 2.55\text{dB} = 1.799$, $G_1 = 14\text{dB} = 25.119$, $F_2 = 2.19\text{dB} = 1.656$, $G_2 = 16\text{dB} = 39.811$, $F_3 = 2.12\text{dB} = 1.629$,
 $F = 1.799 + (1.656 - 1)/25.119 + (1.629 - 1)/25.119/39.811 = 1.826 = 2.614\text{dB}$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ' |
|----|-----------------------------|------------|-------|--------|
| T1 | $-0.320 + j \cdot (0.347)$ | 0.473 | 0.857 | 0.879 |
| T2 | $-0.189 + j \cdot (-0.037)$ | 0.192 | 1.044 | 1.020 |

b) $\mu'(T1) < \mu'(T2)$ so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.038$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 22.182 = 13.460 \text{ dB}$ (L9/2023, S75); However $K = 1.044 > 1$ so maximum gain is $MAG = |S_{21}| / |S_{12}| \cdot [K - \sqrt{K^2 - 1}] = 16.495 = 12.173 \text{ dB}$ (L9/2023, S76)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.087$, $\arg(S_{22}^*) = -140.8^\circ$; $\theta_{S1} = 117.9^\circ$; $\text{Im}(y_S) = -0.175$; **or**
 $\theta_{S2} = 22.9^\circ$; $\text{Im}(y_S) = 0.175$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.595$, $\arg(S_{11}^*) = 155.0^\circ$; $\theta_{S3} = 165.8^\circ$; $\text{Im}(y_S) = -1.481$; **or**
 $\theta_{S4} = 39.2^\circ$; $\text{Im}(y_S) = 1.481$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.175) + (-1.481) = -1.655$; $\theta_{S1} = 117.9^\circ$; $\theta_{P1} = 121.1^\circ$; $\theta_{S3} = 165.8^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.175) + (1.481) = 1.306$; $\theta_{S1} = 117.9^\circ$; $\theta_{P2} = 52.6^\circ$; $\theta_{S4} = 39.2^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.175) + (-1.481) = -1.306$; $\theta_{S2} = 22.9^\circ$; $\theta_{P3} = 127.4^\circ$; $\theta_{S3} = 165.8^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.175) + (1.481) = 1.655$; $\theta_{S2} = 22.9^\circ$; $\theta_{P4} = 58.9^\circ$; $\theta_{S4} = 39.2^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

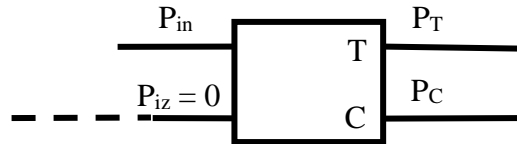
Subject no. 26

1. $z = 0.995 + j \cdot 1.025$; $Y = 1 / 50\Omega / (0.995 + j \cdot 1.025) = 0.0098S + j \cdot (-0.0100)S$; $\Gamma = (z-1)/(z+1) = (0.995 + j \cdot 1.025 - 1)/(0.995 + j \cdot 1.025 + 1) = 0.207 + j \cdot (0.408) = 0.457 \angle 63.1^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.102$, $Z_{0E} = 55.407\Omega$, $Z_{0O} = 45.120\Omega$

b) $P_c = 112.5\mu W = -9.488\text{dBm}$; $P_{in} = P_c + C = -9.488\text{dBm} + 19.8\text{dB} = 10.312\text{dBm} = 10.744 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [\text{mW}] = P_{in} - P_c - P_{iz} = 10.744\text{mW} - 0.1125\text{mW} - 0 = 10.631 \text{ mW} = 10.266 \text{ dBm}$



3. The shunt RL load with $R = 45\Omega$ and $L = 0.507\text{nH}$ has $Z_L = 14.62\Omega + j \cdot (21.08)\Omega$ at 9.8GHz

$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$ (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 3/5) = 0.727$; $\cot(\beta l) = 1.376$; $Z_{in} = 84.02\Omega \angle 61.5^\circ = 40.03\Omega + j \cdot (73.87)\Omega$;

b) If the line becomes short-circuited $Z_L = 0$; $Z_{in} = j \cdot (39.96)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 10\text{dB} + 15\text{dB} + 15\text{dB} = 40\text{dB}$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$F_1 = 2.71\text{dB} = 1.866$, $G_1 = 10\text{dB} = 10.000$, $F_2 = 2.29\text{dB} = 1.694$, $G_2 = 15\text{dB} = 31.623$, $F_3 = 2.39\text{dB} = 1.734$,
 $F = 1.866 + (1.694 - 1)/10.000 + (1.734 - 1)/10.000/31.623 = 1.938 = 2.874\text{dB}$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ |
|----|-----------------------------|------------|-------|-------|
| T1 | $-0.249 + j \cdot (0.415)$ | 0.484 | 0.892 | 0.820 |
| T2 | $-0.169 + j \cdot (-0.018)$ | 0.170 | 1.101 | 1.076 |

b) $\mu (T1) < \mu (T2)$ so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.037$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 21.673 = 13.359 \text{ dB}$ (L9/2023, S75); However $K = 1.101 > 1$ so maximum gain is $MAG = |S_{21}| / |S_{12}| \cdot [K - \sqrt{K^2 - 1}] = 13.884 = 11.425 \text{ dB}$ (L9/2023, S76)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.088$, $\arg(S_{22}^*) = -102.5^\circ$; $\theta_{S1} = 98.8^\circ$; $\text{Im}(y_S) = -0.177$; **or**
 $\theta_{S2} = 3.7^\circ$; $\text{Im}(y_S) = 0.177$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.583$, $\arg(S_{11}^*) = 163.0^\circ$; $\theta_{S3} = 161.3^\circ$; $\text{Im}(y_S) = -1.435$; **or**
 $\theta_{S4} = 35.7^\circ$; $\text{Im}(y_S) = 1.435$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.177) + (-1.435) = -1.612$; $\theta_{S1} = 98.8^\circ$; $\theta_{P1} = 121.8^\circ$; $\theta_{S3} = 161.3^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.177) + (1.435) = 1.258$; $\theta_{S1} = 98.8^\circ$; $\theta_{P2} = 51.5^\circ$; $\theta_{S4} = 35.7^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.177) + (-1.435) = -1.258$; $\theta_{S2} = 3.7^\circ$; $\theta_{P3} = 128.5^\circ$; $\theta_{S3} = 161.3^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.177) + (1.435) = 1.612$; $\theta_{S2} = 3.7^\circ$; $\theta_{P4} = 58.2^\circ$; $\theta_{S4} = 35.7^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

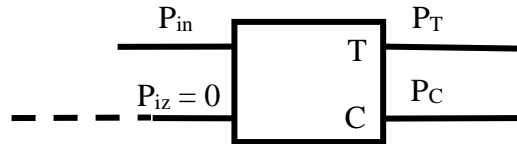
Subject no. 27

1. $z = 0.965 + j \cdot 0.995$; $Y = 1 / 50\Omega / (0.965 + j \cdot 0.995) = 0.0100S + j \cdot (-0.0104)S$; $\Gamma = (z-1)/(z+1) = (0.965 + j \cdot 0.995 - 1)/(0.965 + j \cdot 0.995 + 1) = 0.190 + j \cdot (0.410) = 0.452 \angle 65.2^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.136$, $Z_{0E} = 57.359\Omega$, $Z_{0O} = 43.585\Omega$

b) $P_c = 146.0\mu W = -8.356\text{dBm}$; $P_{in} = P_c + C = -8.356\text{dBm} + 17.3\text{dB} = 8.944\text{dBm} = 7.841 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [\text{mW}] = P_{in} - P_c - P_{iz} = 7.841\text{mW} - 0.1460\text{mW} - 0 = 7.695 \text{ mW} = 8.862 \text{ dBm}$



3. The series RL load with $R = 55\Omega$ and $L = 1.678\text{nH}$ has $Z_L = 55.00\Omega + j \cdot (73.80)\Omega$ at 7.0GHz

$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$ (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 7/8) = -1.000$; $\cot(\beta l) = -1.000$; $Z_{in} = 20.43\Omega \angle 57.4^\circ = 11.02\Omega + j \cdot (17.21)\Omega$;

b) If the line becomes open-circuited $Z_L = \infty$; $Z_{in} = j \cdot (40.00)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 14\text{dB} + 10\text{dB} + 17\text{dB} = 41\text{dB}$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$F_1 = 2.23\text{dB} = 1.671$, $G_1 = 14\text{dB} = 25.119$, $F_2 = 2.18\text{dB} = 1.652$, $G_2 = 10\text{dB} = 10.000$, $F_3 = 2.97\text{dB} = 1.982$,
 $F = 1.671 + (1.652 - 1)/25.119 + (1.982 - 1)/25.119/10.000 = 1.701 = 2.307\text{dB}$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ' |
|----|-----------------------------|------------|-------|--------|
| T1 | $-0.430 + j \cdot (0.204)$ | 0.476 | 0.810 | 0.848 |
| T2 | $-0.223 + j \cdot (-0.098)$ | 0.243 | 0.943 | 0.973 |

b) $\mu'(T1) < \mu'(T2)$ so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.055$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 23.227 = 13.660 \text{ dB}$ (L9/2023, S75)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.110$, $\arg(S_{22}^*) = 160.7^\circ$; $\theta_{S1} = 147.8^\circ$; $\text{Im}(y_S) = -0.221$; **or**
 $\theta_{S2} = 51.5^\circ$; $\text{Im}(y_S) = 0.221$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.619$, $\arg(S_{11}^*) = 139.3^\circ$; $\theta_{S3} = 174.5^\circ$; $\text{Im}(y_S) = -1.576$; **or**
 $\theta_{S4} = 46.2^\circ$; $\text{Im}(y_S) = 1.576$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.221) + (-1.576) = -1.798$; $\theta_{S1} = 147.8^\circ$; $\theta_{P1} = 119.1^\circ$; $\theta_{S3} = 174.5^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.221) + (1.576) = 1.355$; $\theta_{S1} = 147.8^\circ$; $\theta_{P2} = 53.6^\circ$; $\theta_{S4} = 46.2^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.221) + (-1.576) = -1.355$; $\theta_{S2} = 51.5^\circ$; $\theta_{P3} = 126.4^\circ$; $\theta_{S3} = 174.5^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.221) + (1.576) = 1.798$; $\theta_{S2} = 51.5^\circ$; $\theta_{P4} = 60.9^\circ$; $\theta_{S4} = 46.2^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

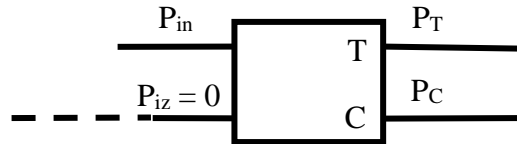
Subject no. 28

1. $z = 1.095 + j \cdot 1.045$; $Y = 1 / 50\Omega / (1.095 + j \cdot 1.045) = 0.0096S + j \cdot (-0.0091)S$; $\Gamma = (z-1)/(z+1) = (1.095 + j \cdot 1.045 - 1)/(1.095 + j \cdot 1.045 + 1) = 0.236 + j \cdot (0.381) = 0.448 \angle 58.3^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.133$, $Z_{0E} = 57.178\Omega$, $Z_{0O} = 43.723\Omega$

b) $P_c = 104.0\mu W = -9.830dBm$; $P_{in} = P_c + C = -9.830dBm + 17.5dB = 7.670dBm = 5.848 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [mW] = P_{in} - P_c - P_{iz} = 5.848mW - 0.1040mW - 0 = 5.744 \text{ mW} = 7.592 \text{ dBm}$



3. The series RL load with $R = 71\Omega$ and $L = 1.417nH$ has $Z_L = 71.00\Omega + j \cdot (66.77)\Omega$ at 7.5GHz

$$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$$
 (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 6/8) = \infty$; $\cot(\beta l) = 0.000$; $Z_{in} = 20.78\Omega \angle -43.2^\circ = 15.13\Omega + j \cdot (-14.23)\Omega$;

b) If the line becomes open-circuited $Z_L = \infty$; $Z_{in} = j \cdot (0.00)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 11dB + 13dB + 18dB = 42dB$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$F_1 = 2.41dB = 1.742$, $G_1 = 11dB = 12.589$, $F_2 = 2.30dB = 1.698$, $G_2 = 13dB = 19.953$, $F_3 = 2.00dB = 1.585$,
 $F = 1.742 + (1.698 - 1)/12.589 + (1.585 - 1)/12.589/19.953 = 1.800 = 2.552dB$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ' |
|----|-----------------------------|------------|-------|--------|
| T1 | $-0.228 + j \cdot (0.435)$ | 0.491 | 0.902 | 0.912 |
| T2 | $-0.164 + j \cdot (-0.014)$ | 0.164 | 1.116 | 1.052 |

b) $\mu'(T1) < \mu'(T2)$ so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.036$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 21.545 = 13.334 \text{ dB}$ (L9/2023, S75); However $K = 1.116 > 1$ so maximum gain is $MAG = |S_{21}| / |S_{12}| \cdot [K - \sqrt{K^2 - 1}] = 13.378 = 11.264 \text{ dB}$ (L9/2023, S76)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.088$, $\arg(S_{22}^*) = -92.9^\circ$; $\theta_{S1} = 94.0^\circ$; $\text{Im}(y_S) = -0.177$; **or**
 $\theta_{S2} = 178.9^\circ$; $\text{Im}(y_S) = 0.177$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.580$, $\arg(S_{11}^*) = 165.0^\circ$; $\theta_{S3} = 160.2^\circ$; $\text{Im}(y_S) = -1.424$; **or**
 $\theta_{S4} = 34.8^\circ$; $\text{Im}(y_S) = 1.424$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.177) + (-1.424) = -1.601$; $\theta_{S1} = 94.0^\circ$; $\theta_{P1} = 122.0^\circ$; $\theta_{S3} = 160.2^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.177) + (1.424) = 1.247$; $\theta_{S1} = 94.0^\circ$; $\theta_{P2} = 51.3^\circ$; $\theta_{S4} = 34.8^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.177) + (-1.424) = -1.247$; $\theta_{S2} = 178.9^\circ$; $\theta_{P3} = 128.7^\circ$; $\theta_{S3} = 160.2^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.177) + (1.424) = 1.601$; $\theta_{S2} = 178.9^\circ$; $\theta_{P4} = 58.0^\circ$; $\theta_{S4} = 34.8^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

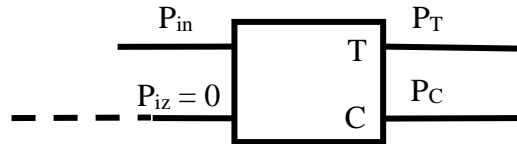
Subject no. 29

1. $z = 0.895 + j \cdot 1.210$; $Y = 1 / 50\Omega / (0.895 + j \cdot 1.210) = 0.0079S + j \cdot (-0.0107)S$; $\Gamma = (z-1)/(z+1) = (0.895 + j \cdot 1.210 - 1)/(0.895 + j \cdot 1.210 + 1) = 0.250 + j \cdot (0.479) = 0.540 \angle 62.4^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.123$, $Z_{0E} = 56.581\Omega$, $Z_{0O} = 44.184\Omega$

b) $P_c = 130.0\mu W = -8.861\text{dBm}$; $P_{in} = P_c + C = -8.861\text{dBm} + 18.2\text{dB} = 9.339\text{dBm} = 8.589 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [\text{mW}] = P_{in} - P_c - P_{iz} = 8.589\text{mW} - 0.1300\text{mW} - 0 = 8.459 \text{ mW} = 9.273 \text{ dBm}$



3. The shunt RC load with $R = 37\Omega$ and $C = 0.310\text{pF}$ has $Z_L = 24.35\Omega + j \cdot (-17.55)\Omega$ at 10.0GHz

$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$ (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 6/8) = \infty$; $\cot(\beta l) = 0.000$; $Z_{in} = 240.70\Omega \angle 35.8^\circ = 195.27\Omega + j \cdot (140.73)\Omega$;

b) If the line becomes short-circuited $Z_L = 0$; $Z_{in} = j \cdot (0.00)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 19\text{dB} + 19\text{dB} + 17\text{dB} = 55\text{dB}$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$F_1 = 2.30\text{dB} = 1.698$, $G_1 = 19\text{dB} = 79.433$, $F_2 = 2.67\text{dB} = 1.849$, $G_2 = 19\text{dB} = 79.433$, $F_3 = 2.50\text{dB} = 1.778$,
 $F = 1.698 + (1.849 - 1)/79.433 + (1.778 - 1)/79.433/79.433 = 1.709 = 2.328\text{dB}$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ' |
|----|-----------------------------|------------|-------|--------|
| T1 | $-0.157 + j \cdot (0.456)$ | 0.483 | 0.912 | 0.923 |
| T2 | $-0.146 + j \cdot (-0.003)$ | 0.146 | 1.146 | 1.065 |

b) $\mu'(\text{T1}) < \mu'(\text{T2})$ so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.048$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 21.109 = 13.245 \text{ dB}$ (L9/2023, S75); However $K = 1.146 > 1$ so maximum gain is $MAG = |S_{21}| / |S_{12}| \cdot [K - \sqrt{K^2 - 1}] = 12.380 = 10.927 \text{ dB}$ (L9/2023, S76)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.118$, $\arg(S_{22}^*) = -80.3^\circ$; $\theta_{S1} = 88.5^\circ$; $\text{Im}(y_S) = -0.238$; **or**
 $\theta_{S2} = 171.8^\circ$; $\text{Im}(y_S) = 0.238$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.577$, $\arg(S_{11}^*) = 171.0^\circ$; $\theta_{S3} = 157.1^\circ$; $\text{Im}(y_S) = -1.413$; **or**
 $\theta_{S4} = 31.9^\circ$; $\text{Im}(y_S) = 1.413$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.238) + (-1.413) = -1.651$; $\theta_{S1} = 88.5^\circ$; $\theta_{P1} = 121.2^\circ$; $\theta_{S3} = 157.1^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.238) + (1.413) = 1.175$; $\theta_{S1} = 88.5^\circ$; $\theta_{P2} = 49.6^\circ$; $\theta_{S4} = 31.9^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.238) + (-1.413) = -1.175$; $\theta_{S2} = 171.8^\circ$; $\theta_{P3} = 130.4^\circ$; $\theta_{S3} = 157.1^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.238) + (1.413) = 1.651$; $\theta_{S2} = 171.8^\circ$; $\theta_{P4} = 58.8^\circ$; $\theta_{S4} = 31.9^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

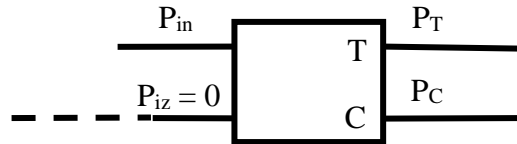
Subject no. 30

1. $z = 1.155 + j \cdot 1.110$; $Y = 1 / 50\Omega / (1.155 + j \cdot 1.110) = 0.0090S + j \cdot (-0.0087)S$; $\Gamma = (z-1)/(z+1) = (1.155 + j \cdot 1.110 - 1)/(1.155 + j \cdot 1.110 + 1) = 0.267 + j \cdot (0.378) = 0.462 \angle 54.8^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.117$, $Z_{0E} = 56.264\Omega$, $Z_{0O} = 44.433\Omega$

b) $P_c = 92.5\mu W = -10.339\text{dBm}$; $P_{in} = P_c + C = -10.339\text{dBm} + 18.6\text{dB} = 8.261\text{dBm} = 6.701 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [\text{mW}] = P_{in} - P_c - P_{iz} = 6.701\text{mW} - 0.0925\text{mW} - 0 = 6.609 \text{ mW} = 8.201 \text{ dBm}$



3. The series RC load with $R = 31\Omega$ and $C = 0.402\text{pF}$ has $Z_L = 31.00\Omega + j \cdot (-59.99)\Omega$ at 6.6GHz

$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$ (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 1/3) = -1.732$; $\cot(\beta l) = -0.577$; $Z_{in} = 202.69\Omega \angle 42.0^\circ = 150.69\Omega + j \cdot (135.55)\Omega$;

b) If the line becomes short-circuited $Z_L = 0$; $Z_{in} = j \cdot (-121.24)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 12\text{dB} + 18\text{dB} + 18\text{dB} = 48\text{dB}$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$F_1 = 2.46\text{dB} = 1.762$, $G_1 = 12\text{dB} = 15.849$, $F_2 = 2.95\text{dB} = 1.972$, $G_2 = 18\text{dB} = 63.096$, $F_3 = 2.59\text{dB} = 1.816$,
 $F = 1.762 + (1.972 - 1)/15.849 + (1.816 - 1)/15.849/63.096 = 1.824 = 2.611\text{dB}$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ |
|----|-----------------------------|------------|-------|-------|
| T1 | $-0.383 + j \cdot (-0.306)$ | 0.490 | 0.577 | 0.455 |
| T2 | $-0.165 + j \cdot (-0.361)$ | 0.397 | 0.687 | 0.746 |

b) $\mu (T1) < \mu (T2)$ so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.300$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 31.000 = 14.914 \text{ dB}$ (L9/2023, S75)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.301$, $\arg(S_{22}^*) = 98.5^\circ$; $\theta_{S1} = 4.5^\circ$; $\text{Im}(y_S) = -0.631$; **or** $\theta_{S2} = 77.0^\circ$; $\text{Im}(y_S) = 0.631$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.742$, $\arg(S_{11}^*) = 94.0^\circ$; $\theta_{S3} = 22.0^\circ$; $\text{Im}(y_S) = -2.214$; **or** $\theta_{S4} = 64.0^\circ$; $\text{Im}(y_S) = 2.214$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.631) + (-2.214) = -2.845$; $\theta_{S1} = 4.5^\circ$; $\theta_{P1} = 109.4^\circ$; $\theta_{S3} = 22.0^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.631) + (2.214) = 1.582$; $\theta_{S1} = 4.5^\circ$; $\theta_{P2} = 57.7^\circ$; $\theta_{S4} = 64.0^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.631) + (-2.214) = -1.582$; $\theta_{S2} = 77.0^\circ$; $\theta_{P3} = 122.3^\circ$; $\theta_{S3} = 22.0^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.631) + (2.214) = 2.845$; $\theta_{S2} = 77.0^\circ$; $\theta_{P4} = 70.6^\circ$; $\theta_{S4} = 64.0^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

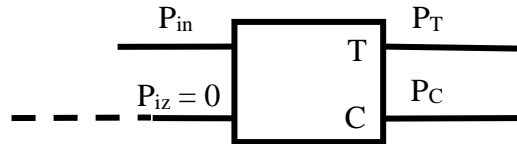
Subject no. 31

1. $z = 1.050 - j \cdot 1.165$; $Y = 1 / 50\Omega / (1.050 - j \cdot 1.165) = 0.0085S + j \cdot (0.0095)S$; $\Gamma = (z-1)/(z+1) = (1.050 - j \cdot 1.165 - 1)/(1.050 - j \cdot 1.165 + 1) = 0.263 + j \cdot (-0.419) = 0.495 \angle -57.9^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.104$, $Z_{0E} = 55.474\Omega$, $Z_{0O} = 45.066\Omega$

b) $P_c = 96.5\mu W = -10.155\text{dBm}$; $P_{in} = P_c + C = -10.155\text{dBm} + 19.7\text{dB} = 9.545\text{dBm} = 9.006 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [\text{mW}] = P_{in} - P_c - P_{iz} = 9.006\text{mW} - 0.0965\text{mW} - 0 = 8.909 \text{ mW} = 9.498 \text{ dBm}$



3. The shunt RC load with $R = 41\Omega$ and $C = 0.566\text{pF}$ has $Z_L = 16.88\Omega + j \cdot (-20.18)\Omega$ at 8.2GHz

$$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$$
 (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 2/6) = -1.732$; $\cot(\beta l) = -0.577$; $Z_{in} = 195.20\Omega \angle -32.9^\circ = 163.98\Omega + j \cdot (-105.90)\Omega$;

b) If the line becomes open-circuited $Z_L = \infty$; $Z_{in} = j \cdot (34.64)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 14\text{dB} + 11\text{dB} + 18\text{dB} = 43\text{dB}$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$F_1 = 2.14\text{dB} = 1.637$, $G_1 = 14\text{dB} = 25.119$, $F_2 = 2.28\text{dB} = 1.690$, $G_2 = 11\text{dB} = 12.589$, $F_3 = 2.92\text{dB} = 1.959$,
 $F = 1.637 + (1.690 - 1)/25.119 + (1.959 - 1)/25.119/12.589 = 1.667 = 2.220\text{dB}$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ |
|----|-----------------------------|------------|-------|-------|
| T1 | $-0.287 + j \cdot (0.381)$ | 0.476 | 0.872 | 0.793 |
| T2 | $-0.179 + j \cdot (-0.026)$ | 0.181 | 1.072 | 1.054 |

b) $\mu (T1) < \mu (T2)$ so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.037$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 21.927 = 13.410 \text{ dB}$ (L9/2023, S75); However $K = 1.072 > 1$ so maximum gain is $MAG = |S_{21}| / |S_{12}| \cdot [K - \sqrt{K^2 - 1}] = 15.041 = 11.773 \text{ dB}$ (L9/2023, S76)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.087$, $\arg(S_{22}^*) = -121.6^\circ$; $\theta_{S1} = 108.3^\circ$; $\text{Im}(y_S) = -0.175$; **or**
 $\theta_{S2} = 13.3^\circ$; $\text{Im}(y_S) = 0.175$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.589$, $\arg(S_{11}^*) = 159.0^\circ$; $\theta_{S3} = 163.5^\circ$; $\text{Im}(y_S) = -1.458$; **or**
 $\theta_{S4} = 37.5^\circ$; $\text{Im}(y_S) = 1.458$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.175) + (-1.458) = -1.632$; $\theta_{S1} = 108.3^\circ$; $\theta_{P1} = 121.5^\circ$; $\theta_{S3} = 163.5^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.175) + (1.458) = 1.283$; $\theta_{S1} = 108.3^\circ$; $\theta_{P2} = 52.1^\circ$; $\theta_{S4} = 37.5^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.175) + (-1.458) = -1.283$; $\theta_{S2} = 13.3^\circ$; $\theta_{P3} = 127.9^\circ$; $\theta_{S3} = 163.5^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.175) + (1.458) = 1.632$; $\theta_{S2} = 13.3^\circ$; $\theta_{P4} = 58.5^\circ$; $\theta_{S4} = 37.5^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

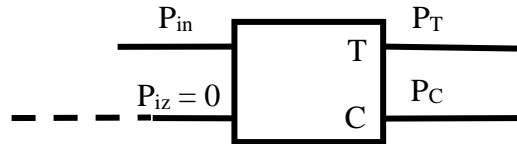
Subject no. 32

1. $z = 1.045 - j \cdot 1.035$; $Y = 1 / 50\Omega / (1.045 - j \cdot 1.035) = 0.0097S + j \cdot (0.0096)S$; $\Gamma = (z-1)/(z+1) = (1.045 - j \cdot 1.035 - 1)/(1.045 - j \cdot 1.035 + 1) = 0.221 + j \cdot (-0.394) = 0.452 \angle -60.7^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.111$, $Z_{0E} = 55.891\Omega$, $Z_{0O} = 44.730\Omega$

b) $P_c = 111.5\mu W = -9.527\text{dBm}$; $P_{in} = P_c + C = -9.527\text{dBm} + 19.1\text{dB} = 9.573\text{dBm} = 9.063 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [\text{mW}] = P_{in} - P_c - P_{iz} = 9.063\text{mW} - 0.1115\text{mW} - 0 = 8.952 \text{ mW} = 9.519 \text{ dBm}$



3. The series RC load with $R = 40\Omega$ and $C = 0.432\text{pF}$ has $Z_L = 40.00\Omega + j \cdot (-51.89)\Omega$ at 7.1GHz

$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$ (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 3/6) = 0.000$; $\cot(\beta l) = \infty$; $Z_{in} = 65.52\Omega \angle -52.4^\circ = 40.00\Omega + j \cdot (-51.89)\Omega$;

b) If the line becomes short-circuited $Z_L = 0$; $Z_{in} = j \cdot (0.00)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 18\text{dB} + 15\text{dB} + 13\text{dB} = 46\text{dB}$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$F_1 = 2.14\text{dB} = 1.637$, $G_1 = 18\text{dB} = 63.096$, $F_2 = 2.92\text{dB} = 1.959$, $G_2 = 15\text{dB} = 31.623$, $F_3 = 2.38\text{dB} = 1.730$,
 $F = 1.637 + (1.959 - 1)/63.096 + (1.730 - 1)/63.096/31.623 = 1.652 = 2.181\text{dB}$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ |
|----|-----------------------------|------------|-------|-------|
| T1 | $-0.446 + j \cdot (-0.200)$ | 0.489 | 0.615 | 0.490 |
| T2 | $-0.203 + j \cdot (-0.297)$ | 0.359 | 0.750 | 0.800 |

b) $\mu (T1) < \mu (T2)$ so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.235$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 28.727 = 14.583 \text{ dB}$ (L9/2023, S75)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.267$, $\arg(S_{22}^*) = 109.6^\circ$; $\theta_{S1} = 178.0^\circ$; $\text{Im}(y_S) = -0.554$; **or**
 $\theta_{S2} = 72.5^\circ$; $\text{Im}(y_S) = 0.554$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.707$, $\arg(S_{11}^*) = 104.0^\circ$; $\theta_{S3} = 15.5^\circ$; $\text{Im}(y_S) = -1.999$; **or**
 $\theta_{S4} = 60.5^\circ$; $\text{Im}(y_S) = 1.999$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.554) + (-1.999) = -2.554$; $\theta_{S1} = 178.0^\circ$; $\theta_{P1} = 111.4^\circ$; $\theta_{S3} = 15.5^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.554) + (1.999) = 1.445$; $\theta_{S1} = 178.0^\circ$; $\theta_{P2} = 55.3^\circ$; $\theta_{S4} = 60.5^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.554) + (-1.999) = -1.445$; $\theta_{S2} = 72.5^\circ$; $\theta_{P3} = 124.7^\circ$; $\theta_{S3} = 15.5^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.554) + (1.999) = 2.554$; $\theta_{S2} = 72.5^\circ$; $\theta_{P4} = 68.6^\circ$; $\theta_{S4} = 60.5^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

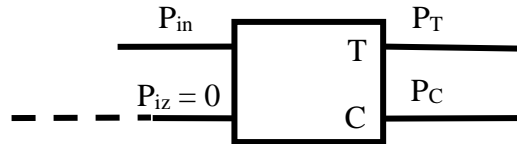
Subject no. 33

1. $z = 1.130 + j \cdot 1.285$; $Y = 1 / 50\Omega / (1.130 + j \cdot 1.285) = 0.0077S + j \cdot (-0.0088)S$; $\Gamma = (z-1)/(z+1) = (1.130 + j \cdot 1.285 - 1)/(1.130 + j \cdot 1.285 + 1) = 0.312 + j \cdot (0.415) = 0.519 \angle 53.1^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.115$, $Z_{0E} = 56.112\Omega$, $Z_{0O} = 44.554\Omega$

b) $P_c = 143.0\mu W = -8.447\text{dBm}$; $P_{in} = P_c + C = -8.447\text{dBm} + 18.8\text{dB} = 10.353\text{dBm} = 10.848 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [\text{mW}] = P_{in} - P_c - P_{iz} = 10.848\text{mW} - 0.1430\text{mW} - 0 = 10.705 \text{ mW} = 10.296 \text{ dBm}$



3. The series RL load with $R = 58\Omega$ and $L = 1.606\text{nH}$ has $Z_L = 58.00\Omega + j \cdot (71.64)\Omega$ at 7.1GHz

$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$ (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 3/6) = 0.000$; $\cot(\beta l) = \infty$; $Z_{in} = 92.18\Omega \angle 51.0^\circ = 58.00\Omega + j \cdot (71.64)\Omega$;

b) If the line becomes open-circuited $Z_L = \infty$; $Z_{in} = j \cdot (0.00)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 19\text{dB} + 14\text{dB} + 12\text{dB} = 45\text{dB}$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$F_1 = 2.05\text{dB} = 1.603$, $G_1 = 19\text{dB} = 79.433$, $F_2 = 2.34\text{dB} = 1.714$, $G_2 = 14\text{dB} = 25.119$, $F_3 = 2.10\text{dB} = 1.622$,

$F = 1.603 + (1.714 - 1)/79.433 + (1.622 - 1)/79.433/25.119 = 1.613 = 2.075\text{dB}$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ' |
|----|-----------------------------|------------|-------|--------|
| T1 | $-0.452 + j \cdot (0.145)$ | 0.475 | 0.779 | 0.828 |
| T2 | $-0.230 + j \cdot (-0.128)$ | 0.263 | 0.908 | 0.957 |

b) $\mu' (T1) < \mu' (T2)$ so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.074$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 23.636 = 13.736 \text{ dB}$ (L9/2023, S75)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.135$, $\arg(S_{22}^*) = 150.2^\circ$; $\theta_{S1} = 153.8^\circ$; $\text{Im}(y_S) = -0.272$; **or** $\theta_{S2} = 56.0^\circ$; $\text{Im}(y_S) = 0.272$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.628$, $\arg(S_{11}^*) = 133.6^\circ$; $\theta_{S3} = 177.7^\circ$; $\text{Im}(y_S) = -1.614$; **or** $\theta_{S4} = 48.7^\circ$; $\text{Im}(y_S) = 1.614$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.272) + (-1.614) = -1.886$; $\theta_{S1} = 153.8^\circ$; $\theta_{P1} = 117.9^\circ$; $\theta_{S3} = 177.7^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.272) + (1.614) = 1.341$; $\theta_{S1} = 153.8^\circ$; $\theta_{P2} = 53.3^\circ$; $\theta_{S4} = 48.7^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.272) + (-1.614) = -1.341$; $\theta_{S2} = 56.0^\circ$; $\theta_{P3} = 126.7^\circ$; $\theta_{S3} = 177.7^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.272) + (1.614) = 1.886$; $\theta_{S2} = 56.0^\circ$; $\theta_{P4} = 62.1^\circ$; $\theta_{S4} = 48.7^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

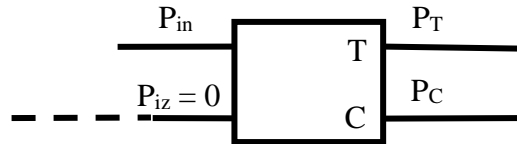
Subject no. 34

1. $z = 1.015 - j \cdot 1.245$; $Y = 1 / 50\Omega / (1.015 - j \cdot 1.245) = 0.0079S + j \cdot (0.0097)S$; $\Gamma = (z-1)/(z+1) = (1.015 - j \cdot 1.245 - 1)/(1.015 - j \cdot 1.245 + 1) = 0.282 + j \cdot (-0.444) = 0.526 \angle -57.6^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.075$, $Z_{0E} = 53.901\Omega$, $Z_{0O} = 46.381\Omega$

b) $P_c = 93.0\mu W = -10.315\text{dBm}$; $P_{in} = P_c + C = -10.315\text{dBm} + 22.5\text{dB} = 12.185\text{dBm} = 16.538 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [\text{mW}] = P_{in} - P_c - P_{iz} = 16.538\text{mW} - 0.0930\text{mW} - 0 = 16.445 \text{ mW} = 12.160 \text{ dBm}$



3. The shunt RC load with $R = 66\Omega$ and $C = 0.383\text{pF}$ has $Z_L = 19.56\Omega + j \cdot (-30.14)\Omega$ at 9.7GHz

$$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$$
 (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 1/3) = -1.732$; $\cot(\beta l) = -0.577$; $Z_{in} = 331.64\Omega \angle -42.1^\circ = 246.01\Omega + j \cdot (-222.42)\Omega$;

b) If the line becomes short-circuited $Z_L = 0$; $Z_{in} = j \cdot (-155.88)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 19\text{dB} + 16\text{dB} + 16\text{dB} = 51\text{dB}$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$F_1 = 2.63\text{dB} = 1.832$, $G_1 = 19\text{dB} = 79.433$, $F_2 = 2.34\text{dB} = 1.714$, $G_2 = 16\text{dB} = 39.811$, $F_3 = 2.51\text{dB} = 1.782$,
 $F = 1.832 + (1.714 - 1)/79.433 + (1.782 - 1)/79.433/39.811 = 1.842 = 2.652\text{dB}$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ |
|----|-----------------------------|------------|-------|-------|
| T1 | $-0.468 + j \cdot (-0.131)$ | 0.486 | 0.645 | 0.519 |
| T2 | $-0.218 + j \cdot (-0.260)$ | 0.339 | 0.783 | 0.827 |

b) $\mu (T1) < \mu (T2)$ so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.193$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 27.412 = 14.379 \text{ dB}$ (L9/2023, S75)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.241$, $\arg(S_{22}^*) = 116.7^\circ$; $\theta_{S1} = 173.6^\circ$; $\text{Im}(y_S) = -0.497$; **or**
 $\theta_{S2} = 69.7^\circ$; $\text{Im}(y_S) = 0.497$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.688$, $\arg(S_{11}^*) = 110.0^\circ$; $\theta_{S3} = 11.7^\circ$; $\text{Im}(y_S) = -1.896$; **or**
 $\theta_{S4} = 58.3^\circ$; $\text{Im}(y_S) = 1.896$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.497) + (-1.896) = -2.393$; $\theta_{S1} = 173.6^\circ$; $\theta_{P1} = 112.7^\circ$; $\theta_{S3} = 11.7^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.497) + (1.896) = 1.399$; $\theta_{S1} = 173.6^\circ$; $\theta_{P2} = 54.5^\circ$; $\theta_{S4} = 58.3^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.497) + (-1.896) = -1.399$; $\theta_{S2} = 69.7^\circ$; $\theta_{P3} = 125.5^\circ$; $\theta_{S3} = 11.7^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.497) + (1.896) = 2.393$; $\theta_{S2} = 69.7^\circ$; $\theta_{P4} = 67.3^\circ$; $\theta_{S4} = 58.3^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

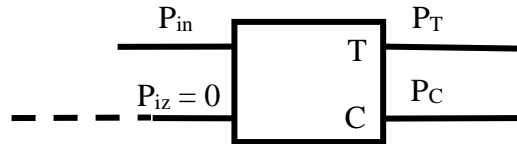
Subject no. 35

1. $z = 0.720 - j \cdot 0.880$; $Y = 1 / 50\Omega / (0.720 - j \cdot 0.880) = 0.0111S + j \cdot (0.0136)S$; $\Gamma = (z-1)/(z+1) = (0.720 - j \cdot 0.880 - 1)/(0.720 - j \cdot 0.880 + 1) = 0.078 + j \cdot (-0.471) = 0.478 \angle -80.6^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.079$, $Z_{0E} = 54.143\Omega$, $Z_{0O} = 46.174\Omega$

b) $P_c = 67.5\mu W = -11.707\text{dBm}$; $P_{in} = P_c + C = -11.707\text{dBm} + 22.0\text{dB} = 10.293\text{dBm} = 10.698 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [\text{mW}] = P_{in} - P_c - P_{iz} = 10.698\text{mW} - 0.0675\text{mW} - 0 = 10.631 \text{ mW} = 10.266 \text{ dBm}$



3. The shunt RL load with $R = 71\Omega$ and $L = 1.063\text{nH}$ has $Z_L = 32.62\Omega + j \cdot (35.38)\Omega$ at 9.8GHz

$$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$$
 (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 4/6) = 1.732$; $\cot(\beta l) = 0.577$; $Z_{in} = 292.50\Omega \angle 21.6^\circ = 272.03\Omega + j \cdot (107.48)\Omega$;

b) If the line becomes open-circuited $Z_L = \infty$; $Z_{in} = j \cdot (-54.85)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 19\text{dB} + 11\text{dB} + 16\text{dB} = 46\text{dB}$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$F_1 = 2.15\text{dB} = 1.641$, $G_1 = 19\text{dB} = 79.433$, $F_2 = 2.91\text{dB} = 1.954$, $G_2 = 11\text{dB} = 12.589$, $F_3 = 2.18\text{dB} = 1.652$,

$F = 1.641 + (1.954 - 1)/79.433 + (1.652 - 1)/79.433/12.589 = 1.653 = 2.183\text{dB}$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ' |
|----|-----------------------------|------------|-------|--------|
| T1 | $-0.480 + j \cdot (-0.009)$ | 0.480 | 0.707 | 0.788 |
| T2 | $-0.231 + j \cdot (-0.204)$ | 0.308 | 0.834 | 0.922 |

b) $\mu'(T1) < \mu'(T2)$ so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.130$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 25.336 = 14.037 \text{ dB}$ (L9/2023, S75)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.195$, $\arg(S_{22}^*) = 128.8^\circ$; $\theta_{S1} = 166.2^\circ$; $\text{Im}(y_S) = -0.398$; **or** $\theta_{S2} = 65.0^\circ$; $\text{Im}(y_S) = 0.398$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.658$, $\arg(S_{11}^*) = 120.0^\circ$; $\theta_{S3} = 5.6^\circ$; $\text{Im}(y_S) = -1.748$; **or** $\theta_{S4} = 54.4^\circ$; $\text{Im}(y_S) = 1.748$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.398) + (-1.748) = -2.145$; $\theta_{S1} = 166.2^\circ$; $\theta_{P1} = 115.0^\circ$; $\theta_{S3} = 5.6^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.398) + (1.748) = 1.350$; $\theta_{S1} = 166.2^\circ$; $\theta_{P2} = 53.5^\circ$; $\theta_{S4} = 54.4^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.398) + (-1.748) = -1.350$; $\theta_{S2} = 65.0^\circ$; $\theta_{P3} = 126.5^\circ$; $\theta_{S3} = 5.6^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.398) + (1.748) = 2.145$; $\theta_{S2} = 65.0^\circ$; $\theta_{P4} = 65.0^\circ$; $\theta_{S4} = 54.4^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

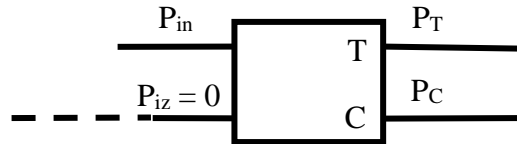
Subject no. 36

1. $z = 0.990 + j \cdot 0.985$; $Y = 1 / 50\Omega / (0.990 + j \cdot 0.985) = 0.0102S + j \cdot (-0.0101)S$; $\Gamma = (z-1)/(z+1) = (0.990 + j \cdot 0.985 - 1)/(0.990 + j \cdot 0.985 + 1) = 0.193 + j \cdot (0.400) = 0.444 \angle 64.2^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.075$, $Z_{0E} = 53.901\Omega$, $Z_{0O} = 46.381\Omega$

b) $P_c = 78.5\mu W = -11.051\text{dBm}$; $P_{in} = P_c + C = -11.051\text{dBm} + 22.5\text{dB} = 11.449\text{dBm} = 13.959 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [\text{mW}] = P_{in} - P_c - P_{iz} = 13.959\text{mW} - 0.0785\text{mW} - 0 = 13.881 \text{ mW} = 11.424 \text{ dBm}$



3. The shunt RC load with $R = 36\Omega$ and $C = 0.331\text{pF}$ has $Z_L = 26.84\Omega + j \cdot (-15.68)\Omega$ at 7.8GHz

$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$ (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 4/6) = 1.732$; $\cot(\beta l) = 0.577$; $Z_{in} = 56.03\Omega \angle 45.0^\circ = 39.62\Omega + j \cdot (39.62)\Omega$;

b) If the line becomes short-circuited $Z_L = 0$; $Z_{in} = j \cdot (103.92)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 11\text{dB} + 12\text{dB} + 18\text{dB} = 41\text{dB}$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$F_1 = 2.21\text{dB} = 1.663$, $G_1 = 11\text{dB} = 12.589$, $F_2 = 2.62\text{dB} = 1.828$, $G_2 = 12\text{dB} = 15.849$, $F_3 = 2.02\text{dB} = 1.592$,
 $F = 1.663 + (1.828 - 1)/12.589 + (1.592 - 1)/12.589/15.849 = 1.732 = 2.386\text{dB}$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ |
|----|-----------------------------|------------|-------|-------|
| T1 | $-0.436 + j \cdot (-0.224)$ | 0.490 | 0.607 | 0.482 |
| T2 | $-0.197 + j \cdot (-0.309)$ | 0.367 | 0.737 | 0.789 |

b) $\mu (T1) < \mu (T2)$ so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.248$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 29.163 = 14.648 \text{ dB}$ (L9/2023, S75)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.274$, $\arg(S_{22}^*) = 107.4^\circ$; $\theta_{S1} = 179.3^\circ$; $\text{Im}(y_S) = -0.570$; **or**
 $\theta_{S2} = 73.4^\circ$; $\text{Im}(y_S) = 0.570$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.714$, $\arg(S_{11}^*) = 102.0^\circ$; $\theta_{S3} = 16.8^\circ$; $\text{Im}(y_S) = -2.040$; **or**
 $\theta_{S4} = 61.2^\circ$; $\text{Im}(y_S) = 2.040$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.570) + (-2.040) = -2.609$; $\theta_{S1} = 179.3^\circ$; $\theta_{P1} = 111.0^\circ$; $\theta_{S3} = 16.8^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.570) + (2.040) = 1.470$; $\theta_{S1} = 179.3^\circ$; $\theta_{P2} = 55.8^\circ$; $\theta_{S4} = 61.2^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.570) + (-2.040) = -1.470$; $\theta_{S2} = 73.4^\circ$; $\theta_{P3} = 124.2^\circ$; $\theta_{S3} = 16.8^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.570) + (2.040) = 2.609$; $\theta_{S2} = 73.4^\circ$; $\theta_{P4} = 69.0^\circ$; $\theta_{S4} = 61.2^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

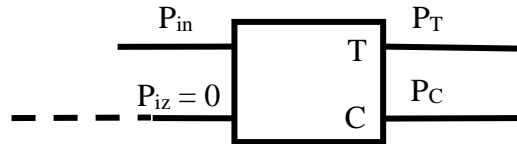
Subject no. 37

1. $z = 0.755 + j \cdot 0.705$; $Y = 1 / 50\Omega / (0.755 + j \cdot 0.705) = 0.0142S + j \cdot (-0.0132)S$; $\Gamma = (z-1)/(z+1) = (0.755 + j \cdot 0.705 - 1)/(0.755 + j \cdot 0.705 + 1) = 0.019 + j \cdot (0.394) = 0.395 \angle 87.3^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.110$, $Z_{0E} = 55.819\Omega$, $Z_{0O} = 44.788\Omega$

b) $P_c = 64.5\mu W = -11.904\text{dBm}$; $P_{in} = P_c + C = -11.904\text{dBm} + 19.2\text{dB} = 7.296\text{dBm} = 5.365 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [\text{mW}] = P_{in} - P_c - P_{iz} = 5.365\text{mW} - 0.0645\text{mW} - 0 = 5.300 \text{ mW} = 7.243 \text{ dBm}$



3. The series RC load with $R = 29\Omega$ and $C = 0.426\text{pF}$ has $Z_L = 29.00\Omega + j \cdot (-47.29)\Omega$ at 7.9GHz

$$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$$
 (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 4/6) = 1.732$; $\cot(\beta l) = 0.577$; $Z_{in} = 8.78\Omega \angle 1.4^\circ = 8.78\Omega + j \cdot (0.22)\Omega$;

b) If the line becomes short-circuited $Z_L = 0$; $Z_{in} = j \cdot (60.62)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 16\text{dB} + 15\text{dB} + 11\text{dB} = 42\text{dB}$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$F_1 = 2.39\text{dB} = 1.734$, $G_1 = 16\text{dB} = 39.811$, $F_2 = 2.09\text{dB} = 1.618$, $G_2 = 15\text{dB} = 31.623$, $F_3 = 2.60\text{dB} = 1.820$,
 $F = 1.734 + (1.618 - 1)/39.811 + (1.820 - 1)/39.811/31.623 = 1.750 = 2.430\text{dB}$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ' |
|----|-----------------------------|------------|-------|--------|
| T1 | $-0.479 + j \cdot (-0.034)$ | 0.480 | 0.694 | 0.781 |
| T2 | $-0.229 + j \cdot (-0.214)$ | 0.314 | 0.824 | 0.917 |

b) $\mu'(T1) < \mu'(T2)$ so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.141$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 25.736 = 14.105 \text{ dB}$ (L9/2023, S75)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.204$, $\arg(S_{22}^*) = 126.4^\circ$; $\theta_{S1} = 167.7^\circ$; $\text{Im}(y_S) = -0.417$; **or** $\theta_{S2} = 65.9^\circ$; $\text{Im}(y_S) = 0.417$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.664$, $\arg(S_{11}^*) = 118.0^\circ$; $\theta_{S3} = 6.8^\circ$; $\text{Im}(y_S) = -1.776$; **or** $\theta_{S4} = 55.2^\circ$; $\text{Im}(y_S) = 1.776$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.417) + (-1.776) = -2.193$; $\theta_{S1} = 167.7^\circ$; $\theta_{P1} = 114.5^\circ$; $\theta_{S3} = 6.8^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.417) + (1.776) = 1.359$; $\theta_{S1} = 167.7^\circ$; $\theta_{P2} = 53.7^\circ$; $\theta_{S4} = 55.2^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.417) + (-1.776) = -1.359$; $\theta_{S2} = 65.9^\circ$; $\theta_{P3} = 126.3^\circ$; $\theta_{S3} = 6.8^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.417) + (1.776) = 2.193$; $\theta_{S2} = 65.9^\circ$; $\theta_{P4} = 65.5^\circ$; $\theta_{S4} = 55.2^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

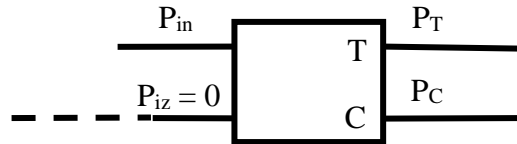
Subject no. 38

1. $z = 1.250 - j \cdot 1.275$; $Y = 1 / 50\Omega / (1.250 - j \cdot 1.275) = 0.0078S + j \cdot (0.0080)S$; $\Gamma = (z-1)/(z+1) = (1.250 - j \cdot 1.275 - 1)/(1.250 - j \cdot 1.275 + 1) = 0.327 + j \cdot (-0.381) = 0.502 \angle -49.4^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.101$, $Z_{0E} = 55.342\Omega$, $Z_{0O} = 45.174\Omega$

b) $P_c = 85.5\mu W = -10.680\text{dBm}$; $P_{in} = P_c + C = -10.680\text{dBm} + 19.9\text{dB} = 9.220\text{dBm} = 8.355 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [\text{mW}] = P_{in} - P_c - P_{iz} = 8.355\text{mW} - 0.0855\text{mW} - 0 = 8.270 \text{ mW} = 9.175 \text{ dBm}$



3. The shunt RL load with $R = 55\Omega$ and $L = 1.165\text{nH}$ has $Z_L = 27.08\Omega + j \cdot (27.50)\Omega$ at 7.4GHz

$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$ (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 1/6) = 1.732$; $\cot(\beta l) = 0.577$; $Z_{in} = 84.59\Omega \angle -24.9^\circ = 76.75\Omega + j \cdot (-35.57)\Omega$;

b) If the line becomes short-circuited $Z_L = 0$; $Z_{in} = j \cdot (69.28)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 13\text{dB} + 15\text{dB} + 15\text{dB} = 43\text{dB}$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$F_1 = 2.18\text{dB} = 1.652$, $G_1 = 13\text{dB} = 19.953$, $F_2 = 2.70\text{dB} = 1.862$, $G_2 = 15\text{dB} = 31.623$, $F_3 = 2.24\text{dB} = 1.675$,
 $F = 1.652 + (1.862 - 1)/19.953 + (1.675 - 1)/19.953/31.623 = 1.696 = 2.295\text{dB}$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ' |
|----|----------------------------|------------|-------|--------|
| T1 | $-0.110 + j \cdot (0.469)$ | 0.482 | 0.923 | 0.933 |
| T2 | $-0.134 + j \cdot (0.002)$ | 0.134 | 1.168 | 1.073 |

b) $\mu'(T1) < \mu'(T2)$ so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.055$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 20.818 = 13.184 \text{ dB}$ (L9/2023, S75); However $K = 1.168 > 1$ so maximum gain is $MAG = |S_{21}| / |S_{12}| \cdot [K - \sqrt{K^2 - 1}] = 11.761 = 10.705 \text{ dB}$ (L9/2023, S76)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.138$, $\arg(S_{22}^*) = -71.9^\circ$; $\theta_{S1} = 84.9^\circ$; $\text{Im}(y_S) = -0.279$; **or**
 $\theta_{S2} = 167.0^\circ$; $\text{Im}(y_S) = 0.279$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.575$, $\arg(S_{11}^*) = 175.0^\circ$; $\theta_{S3} = 155.0^\circ$; $\text{Im}(y_S) = -1.406$; **or**
 $\theta_{S4} = 30.0^\circ$; $\text{Im}(y_S) = 1.406$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.279) + (-1.406) = -1.684$; $\theta_{S1} = 84.9^\circ$; $\theta_{P1} = 120.7^\circ$; $\theta_{S3} = 155.0^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.279) + (1.406) = 1.127$; $\theta_{S1} = 84.9^\circ$; $\theta_{P2} = 48.4^\circ$; $\theta_{S4} = 30.0^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.279) + (-1.406) = -1.127$; $\theta_{S2} = 167.0^\circ$; $\theta_{P3} = 131.6^\circ$; $\theta_{S3} = 155.0^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.279) + (1.406) = 1.684$; $\theta_{S2} = 167.0^\circ$; $\theta_{P4} = 59.3^\circ$; $\theta_{S4} = 30.0^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

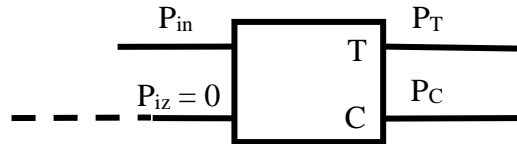
Subject no. 39

1. $z = 1.205 + j \cdot 1.255$; $Y = 1 / 50\Omega / (1.205 + j \cdot 1.255) = 0.0080S + j \cdot (-0.0083)S$; $\Gamma = (z-1)/(z+1) = (1.205 + j \cdot 1.255 - 1)/(1.205 + j \cdot 1.255 + 1) = 0.315 + j \cdot (0.390) = 0.501 \angle 51.1^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.117$, $Z_{0E} = 56.264\Omega$, $Z_{0O} = 44.433\Omega$

b) $P_c = 137.0\mu W = -8.633dBm$; $P_{in} = P_c + C = -8.633dBm + 18.6dB = 9.967dBm = 9.925 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [mW] = P_{in} - P_c - P_{iz} = 9.925mW - 0.1370mW - 0 = 9.788 \text{ mW} = 9.907 \text{ dBm}$



3. The shunt RC load with $R = 64\Omega$ and $C = 0.253pF$ has $Z_L = 33.09\Omega + j \cdot (-31.98)\Omega$ at 9.5GHz

$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$ (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 4/5) = -3.078$; $\cot(\beta l) = -0.325$; $Z_{in} = 54.03\Omega \angle 41.9^\circ = 40.22\Omega + j \cdot (36.07)\Omega$;

b) If the line becomes short-circuited $Z_L = 0$; $Z_{in} = j \cdot (-123.11)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 14dB + 12dB + 15dB = 41dB$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$F_1 = 2.76dB = 1.888$, $G_1 = 14dB = 25.119$, $F_2 = 2.86dB = 1.932$, $G_2 = 12dB = 15.849$, $F_3 = 2.67dB = 1.849$,
 $F = 1.888 + (1.932 - 1)/25.119 + (1.849 - 1)/25.119/15.849 = 1.927 = 2.849dB$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ' |
|----|-----------------------------|------------|-------|--------|
| T1 | $-0.478 + j \cdot (-0.058)$ | 0.481 | 0.682 | 0.775 |
| T2 | $-0.227 + j \cdot (-0.225)$ | 0.320 | 0.813 | 0.913 |

b) $\mu'(T1) < \mu'(T2)$ so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.153$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 26.143 = 14.174 \text{ dB}$ (L9/2023, S75)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.214$, $\arg(S_{22}^*) = 123.9^\circ$; $\theta_{S1} = 169.2^\circ$; $\text{Im}(y_S) = -0.438$; **or** $\theta_{S2} = 66.8^\circ$; $\text{Im}(y_S) = 0.438$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.670$, $\arg(S_{11}^*) = 116.0^\circ$; $\theta_{S3} = 8.0^\circ$; $\text{Im}(y_S) = -1.805$; **or** $\theta_{S4} = 56.0^\circ$; $\text{Im}(y_S) = 1.805$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.438) + (-1.805) = -2.243$; $\theta_{S1} = 169.2^\circ$; $\theta_{P1} = 114.0^\circ$; $\theta_{S3} = 8.0^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.438) + (1.805) = 1.367$; $\theta_{S1} = 169.2^\circ$; $\theta_{P2} = 53.8^\circ$; $\theta_{S4} = 56.0^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.438) + (-1.805) = -1.367$; $\theta_{S2} = 66.8^\circ$; $\theta_{P3} = 126.2^\circ$; $\theta_{S3} = 8.0^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.438) + (1.805) = 2.243$; $\theta_{S2} = 66.8^\circ$; $\theta_{P4} = 66.0^\circ$; $\theta_{S4} = 56.0^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

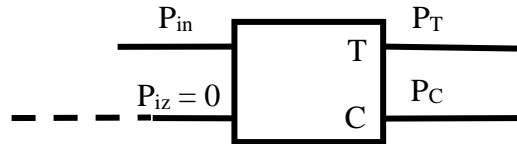
Subject no. 40

1. $z = 0.765 - j \cdot 0.985$; $Y = 1 / 50\Omega / (0.765 - j \cdot 0.985) = 0.0098S + j \cdot (0.0127)S$; $\Gamma = (z-1)/(z+1) = (0.765 - j \cdot 0.985 - 1)/(0.765 - j \cdot 0.985 + 1) = 0.136 + j \cdot (-0.482) = 0.501 \angle -74.3^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.085$, $Z_{0E} = 54.453\Omega$, $Z_{0O} = 45.911\Omega$

b) $P_c = 73.0\mu W = -11.367\text{dBm}$; $P_{in} = P_c + C = -11.367\text{dBm} + 21.4\text{dB} = 10.033\text{dBm} = 10.077 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [\text{mW}] = P_{in} - P_c - P_{iz} = 10.077\text{mW} - 0.0730\text{mW} - 0 = 10.004 \text{ mW} = 10.002 \text{ dBm}$



3. The shunt RC load with $R = 46\Omega$ and $C = 0.738\text{pF}$ has $Z_L = 10.73\Omega + j \cdot (-19.45)\Omega$ at 8.5GHz

$$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$$
 (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 2/3) = 1.732$; $\cot(\beta l) = 0.577$; $Z_{in} = 39.66\Omega \angle 68.4^\circ = 14.60\Omega + j \cdot (36.87)\Omega$;

b) If the line becomes open-circuited $Z_L = \infty$; $Z_{in} = j \cdot (-28.87)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 17\text{dB} + 16\text{dB} + 17\text{dB} = 50\text{dB}$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$F_1 = 2.38\text{dB} = 1.730$, $G_1 = 17\text{dB} = 50.119$, $F_2 = 2.54\text{dB} = 1.795$, $G_2 = 16\text{dB} = 39.811$, $F_3 = 2.75\text{dB} = 1.884$,
 $F = 1.730 + (1.795 - 1)/50.119 + (1.884 - 1)/50.119/39.811 = 1.746 = 2.421\text{dB}$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ |
|----|-----------------------------|------------|-------|-------|
| T1 | $-0.350 + j \cdot (0.314)$ | 0.470 | 0.849 | 0.763 |
| T2 | $-0.199 + j \cdot (-0.049)$ | 0.205 | 1.018 | 1.013 |

b) $\mu (T1) < \mu (T2)$ so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.039$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 22.436 = 13.510 \text{ dB}$ (L9/2023, S75); However $K = 1.018 > 1$ so maximum gain is $MAG = |S_{21}| / |S_{12}| \cdot [K - \sqrt{K^2 - 1}] = 18.586 = 12.692 \text{ dB}$ (L9/2023, S76)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.087$, $\arg(S_{22}^*) = -160.0^\circ$; $\theta_{S1} = 127.5^\circ$; $\text{Im}(y_S) = -0.175$; **or**
 $\theta_{S2} = 32.5^\circ$; $\text{Im}(y_S) = 0.175$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.601$, $\arg(S_{11}^*) = 151.0^\circ$; $\theta_{S3} = 168.0^\circ$; $\text{Im}(y_S) = -1.504$; **or**
 $\theta_{S4} = 41.0^\circ$; $\text{Im}(y_S) = 1.504$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.175) + (-1.504) = -1.679$; $\theta_{S1} = 127.5^\circ$; $\theta_{P1} = 120.8^\circ$; $\theta_{S3} = 168.0^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.175) + (1.504) = 1.329$; $\theta_{S1} = 127.5^\circ$; $\theta_{P2} = 53.0^\circ$; $\theta_{S4} = 41.0^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.175) + (-1.504) = -1.329$; $\theta_{S2} = 32.5^\circ$; $\theta_{P3} = 127.0^\circ$; $\theta_{S3} = 168.0^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.175) + (1.504) = 1.679$; $\theta_{S2} = 32.5^\circ$; $\theta_{P4} = 59.2^\circ$; $\theta_{S4} = 41.0^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

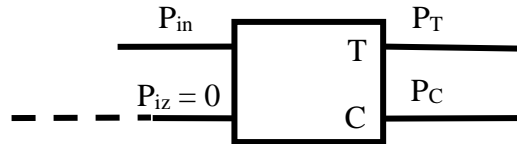
Subject no. 41

1. $z = 1.120 + j \cdot 1.275$; $Y = 1 / 50\Omega / (1.120 + j \cdot 1.275) = 0.0078S + j \cdot (-0.0089)S$; $\Gamma = (z-1)/(z+1) = (1.120 + j \cdot 1.275 - 1)/(1.120 + j \cdot 1.275 + 1) = 0.307 + j \cdot (0.417) = 0.518 \angle 53.6^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.094$, $Z_{0E} = 54.966\Omega$, $Z_{0O} = 45.483\Omega$

b) $P_c = 140.5\mu W = -8.523dBm$; $P_{in} = P_c + C = -8.523dBm + 20.5dB = 11.977dBm = 15.764 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [mW] = P_{in} - P_c - P_{iz} = 15.764mW - 0.1405mW - 0 = 15.624 \text{ mW} = 11.938 \text{ dBm}$



3. The series RC load with $R = 63\Omega$ and $C = 0.279pF$ has $Z_L = 63.00\Omega + j \cdot (-73.13)\Omega$ at 7.8GHz

$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$ (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 6/8) = \infty$; $\cot(\beta l) = 0.000$; $Z_{in} = 25.90\Omega \angle 49.3^\circ = 16.90\Omega + j \cdot (19.62)\Omega$;

b) If the line becomes open-circuited $Z_L = \infty$; $Z_{in} = j \cdot (0.00)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 14dB + 13dB + 17dB = 44dB$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$F_1 = 2.84dB = 1.923$, $G_1 = 14dB = 25.119$, $F_2 = 2.21dB = 1.663$, $G_2 = 13dB = 19.953$, $F_3 = 2.03dB = 1.596$,
 $F = 1.923 + (1.663 - 1)/25.119 + (1.596 - 1)/25.119/19.953 = 1.951 = 2.902dB$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ' |
|----|-----------------------------|------------|-------|--------|
| T1 | $-0.410 + j \cdot (0.244)$ | 0.477 | 0.834 | 0.863 |
| T2 | $-0.216 + j \cdot (-0.080)$ | 0.231 | 0.968 | 0.985 |

b) $\mu' (T1) < \mu' (T2)$ so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.045$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 22.955 = 13.609 \text{ dB}$ (L9/2023, S75)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.094$, $\arg(S_{22}^*) = 167.8^\circ$; $\theta_{S1} = 143.8^\circ$; $\text{Im}(y_S) = -0.189$; **or**
 $\theta_{S2} = 48.4^\circ$; $\text{Im}(y_S) = 0.189$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.613$, $\arg(S_{11}^*) = 143.1^\circ$; $\theta_{S3} = 172.4^\circ$; $\text{Im}(y_S) = -1.552$; **or**
 $\theta_{S4} = 44.5^\circ$; $\text{Im}(y_S) = 1.552$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.189) + (-1.552) = -1.741$; $\theta_{S1} = 143.8^\circ$; $\theta_{P1} = 119.9^\circ$; $\theta_{S3} = 172.4^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.189) + (1.552) = 1.363$; $\theta_{S1} = 143.8^\circ$; $\theta_{P2} = 53.7^\circ$; $\theta_{S4} = 44.5^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.189) + (-1.552) = -1.363$; $\theta_{S2} = 48.4^\circ$; $\theta_{P3} = 126.3^\circ$; $\theta_{S3} = 172.4^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.189) + (1.552) = 1.741$; $\theta_{S2} = 48.4^\circ$; $\theta_{P4} = 60.1^\circ$; $\theta_{S4} = 44.5^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

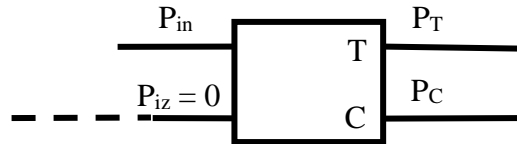
Subject no. 42

1. $z = 1.225 + j \cdot 0.750$; $Y = 1 / 50\Omega / (1.225 + j \cdot 0.750) = 0.0119S + j \cdot (-0.0073)S$; $\Gamma = (z-1)/(z+1) = (1.225 + j \cdot 0.750 - 1)/(1.225 + j \cdot 0.750 + 1) = 0.193 + j \cdot (0.272) = 0.333 \angle 54.7^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.122$, $Z_{0E} = 56.500\Omega$, $Z_{0O} = 44.248\Omega$

b) $P_c = 148.5\mu W = -8.283dBm$; $P_{in} = P_c + C = -8.283dBm + 18.3dB = 10.017dBm = 10.040 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [mW] = P_{in} - P_c - P_{iz} = 10.040mW - 0.1485mW - 0 = 9.891 \text{ mW} = 9.953 \text{ dBm}$



3. The shunt RC load with $R = 29\Omega$ and $C = 0.566pF$ has $Z_L = 16.23\Omega + j \cdot (-14.40)\Omega$ at 8.6GHz

$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$ (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 4/6) = 1.732$; $\cot(\beta l) = 0.577$; $Z_{in} = 116.47\Omega \angle 70.6^\circ = 38.61\Omega + j \cdot (109.88)\Omega$;

b) If the line becomes short-circuited $Z_L = 0$; $Z_{in} = j \cdot (164.54)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 13dB + 13dB + 16dB = 42dB$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$F_1 = 2.89dB = 1.945$, $G_1 = 13dB = 19.953$, $F_2 = 2.88dB = 1.941$, $G_2 = 13dB = 19.953$, $F_3 = 2.58dB = 1.811$,
 $F = 1.945 + (1.941 - 1)/19.953 + (1.811 - 1)/19.953/19.953 = 1.995 = 2.998dB$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ |
|----|-----------------------------|------------|-------|-------|
| T1 | $-0.412 + j \cdot (-0.265)$ | 0.490 | 0.591 | 0.467 |
| T2 | $-0.182 + j \cdot (-0.335)$ | 0.381 | 0.712 | 0.768 |

b) $\mu(T1) < \mu(T2)$ so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.274$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 30.063 = 14.780 \text{ dB}$ (L9/2023, S75)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.288$, $\arg(S_{22}^*) = 102.9^\circ$; $\theta_{S1} = 1.9^\circ$; $\text{Im}(y_S) = -0.601$; **or** $\theta_{S2} = 75.2^\circ$; $\text{Im}(y_S) = 0.601$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.728$, $\arg(S_{11}^*) = 98.0^\circ$; $\theta_{S3} = 19.4^\circ$; $\text{Im}(y_S) = -2.124$; **or** $\theta_{S4} = 62.6^\circ$; $\text{Im}(y_S) = 2.124$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.601) + (-2.124) = -2.725$; $\theta_{S1} = 1.9^\circ$; $\theta_{P1} = 110.2^\circ$; $\theta_{S3} = 19.4^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.601) + (2.124) = 1.522$; $\theta_{S1} = 1.9^\circ$; $\theta_{P2} = 56.7^\circ$; $\theta_{S4} = 62.6^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.601) + (-2.124) = -1.522$; $\theta_{S2} = 75.2^\circ$; $\theta_{P3} = 123.3^\circ$; $\theta_{S3} = 19.4^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.601) + (2.124) = 2.725$; $\theta_{S2} = 75.2^\circ$; $\theta_{P4} = 69.8^\circ$; $\theta_{S4} = 62.6^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

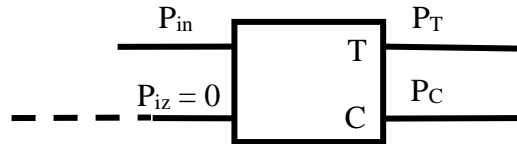
Subject no. 43

1. $z = 1.275 - j \cdot 1.055$; $Y = 1 / 50\Omega / (1.275 - j \cdot 1.055) = 0.0093S + j \cdot (0.0077)S$; $\Gamma = (z-1)/(z+1) = (1.275 - j \cdot 1.055 - 1)/(1.275 - j \cdot 1.055 + 1) = 0.276 + j \cdot (-0.336) = 0.435 \angle -50.5^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.095$, $Z_{0E} = 55.026\Omega$, $Z_{0O} = 45.433\Omega$

b) $P_c = 63.0\mu W = -12.007\text{dBm}$; $P_{in} = P_c + C = -12.007\text{dBm} + 20.4\text{dB} = 8.393\text{dBm} = 6.908 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [\text{mW}] = P_{in} - P_c - P_{iz} = 6.908\text{mW} - 0.0630\text{mW} - 0 = 6.845 \text{ mW} = 8.354 \text{ dBm}$



3. The shunt RC load with $R = 53\Omega$ and $C = 0.466\text{pF}$ has $Z_L = 18.50\Omega + j \cdot (-25.26)\Omega$ at 8.8GHz

$$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$$
 (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 2/6) = -1.732$; $\cot(\beta l) = -0.577$; $Z_{in} = 173.66\Omega \angle -1.6^\circ = 173.59\Omega + j \cdot (-4.95)\Omega$;

b) If the line becomes open-circuited $Z_L = \infty$; $Z_{in} = j \cdot (28.87)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 16\text{dB} + 18\text{dB} + 10\text{dB} = 44\text{dB}$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$F_1 = 2.25\text{dB} = 1.679$, $G_1 = 16\text{dB} = 39.811$, $F_2 = 2.45\text{dB} = 1.758$, $G_2 = 18\text{dB} = 63.096$, $F_3 = 2.48\text{dB} = 1.770$,
 $F = 1.679 + (1.758 - 1)/39.811 + (1.770 - 1)/39.811/63.096 = 1.698 = 2.300\text{dB}$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ' |
|----|----------------------------|------------|-------|--------|
| T1 | $-0.062 + j \cdot (0.484)$ | 0.488 | 0.934 | 0.942 |
| T2 | $-0.123 + j \cdot (0.006)$ | 0.123 | 1.191 | 1.082 |

b) $\mu'(T1) < \mu'(T2)$ so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.063$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 20.527 = 13.123 \text{ dB}$ (L9/2023, S75); However $K = 1.191 > 1$ so maximum gain is $MAG = |S_{21}| / |S_{12}| \cdot [K - \sqrt{K^2 - 1}] = 11.176 = 10.483 \text{ dB}$ (L9/2023, S76)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.159$, $\arg(S_{22}^*) = -63.5^\circ$; $\theta_{S1} = 81.3^\circ$; $\text{Im}(y_S) = -0.322$; **or**
 $\theta_{S2} = 162.2^\circ$; $\text{Im}(y_S) = 0.322$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.573$, $\arg(S_{11}^*) = 179.0^\circ$; $\theta_{S3} = 153.0^\circ$; $\text{Im}(y_S) = -1.398$; **or**
 $\theta_{S4} = 28.0^\circ$; $\text{Im}(y_S) = 1.398$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.322) + (-1.398) = -1.720$; $\theta_{S1} = 81.3^\circ$; $\theta_{P1} = 120.2^\circ$; $\theta_{S3} = 153.0^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.322) + (1.398) = 1.076$; $\theta_{S1} = 81.3^\circ$; $\theta_{P2} = 47.1^\circ$; $\theta_{S4} = 28.0^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.322) + (-1.398) = -1.076$; $\theta_{S2} = 162.2^\circ$; $\theta_{P3} = 132.9^\circ$; $\theta_{S3} = 153.0^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.322) + (1.398) = 1.720$; $\theta_{S2} = 162.2^\circ$; $\theta_{P4} = 59.8^\circ$; $\theta_{S4} = 28.0^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

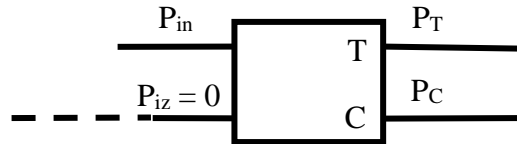
Subject no. 44

1. $z = 0.965 + j \cdot 1.110$; $Y = 1 / 50\Omega / (0.965 + j \cdot 1.110) = 0.0089S + j \cdot (-0.0103)S$; $\Gamma = (z-1)/(z+1) = (0.965 + j \cdot 1.110 - 1)/(0.965 + j \cdot 1.110 + 1) = 0.228 + j \cdot (0.436) = 0.492 \angle 62.3^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.072$, $Z_{0E} = 53.719\Omega$, $Z_{0O} = 46.539\Omega$

b) $P_c = 145.0\mu W = -8.386\text{dBm}$; $P_{in} = P_c + C = -8.386\text{dBm} + 22.9\text{dB} = 14.514\text{dBm} = 28.273 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [\text{mW}] = P_{in} - P_c - P_{iz} = 28.273\text{mW} - 0.1450\text{mW} - 0 = 28.128 \text{ mW} = 14.491 \text{ dBm}$



3. The series RC load with $R = 39\Omega$ and $C = 0.336\text{pF}$ has $Z_L = 39.00\Omega + j \cdot (-48.83)\Omega$ at 9.7GHz

$$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$$
 (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 3/6) = 0.000$; $\cot(\beta l) = \infty$; $Z_{in} = 62.49\Omega \angle -51.4^\circ = 39.00\Omega + j \cdot (-48.83)\Omega$;

b) If the line becomes open-circuited $Z_L = \infty$; $Z_{in} = j \cdot (0.00)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 10\text{dB} + 16\text{dB} + 14\text{dB} = 40\text{dB}$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$F_1 = 2.95\text{dB} = 1.972$, $G_1 = 10\text{dB} = 10.000$, $F_2 = 2.43\text{dB} = 1.750$, $G_2 = 16\text{dB} = 39.811$, $F_3 = 2.34\text{dB} = 1.714$,
 $F = 1.972 + (1.750 - 1)/10.000 + (1.714 - 1)/10.000/39.811 = 2.049 = 3.116\text{dB}$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ' |
|----|----------------------------|------------|-------|--------|
| T1 | $-0.087 + j \cdot (0.478)$ | 0.486 | 0.926 | 0.935 |
| T2 | $-0.129 + j \cdot (0.004)$ | 0.129 | 1.179 | 1.078 |

b) $\mu'(T1) < \mu'(T2)$ so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.060$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 20.673 = 13.154 \text{ dB}$ (L9/2023, S75); However $K = 1.179 > 1$ so maximum gain is $MAG = |S_{21}| / |S_{12}| \cdot [K - \sqrt{K^2 - 1}] = 11.465 = 10.594 \text{ dB}$ (L9/2023, S76)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.149$, $\arg(S_{22}^*) = -67.7^\circ$; $\theta_{S1} = 83.1^\circ$; $\text{Im}(y_S) = -0.301$; **or**
 $\theta_{S2} = 164.6^\circ$; $\text{Im}(y_S) = 0.301$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.574$, $\arg(S_{11}^*) = 177.0^\circ$; $\theta_{S3} = 154.0^\circ$; $\text{Im}(y_S) = -1.402$; **or**
 $\theta_{S4} = 29.0^\circ$; $\text{Im}(y_S) = 1.402$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.301) + (-1.402) = -1.703$; $\theta_{S1} = 83.1^\circ$; $\theta_{P1} = 120.4^\circ$; $\theta_{S3} = 154.0^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.301) + (1.402) = 1.101$; $\theta_{S1} = 83.1^\circ$; $\theta_{P2} = 47.7^\circ$; $\theta_{S4} = 29.0^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.301) + (-1.402) = -1.101$; $\theta_{S2} = 164.6^\circ$; $\theta_{P3} = 132.3^\circ$; $\theta_{S3} = 154.0^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.301) + (1.402) = 1.703$; $\theta_{S2} = 164.6^\circ$; $\theta_{P4} = 59.6^\circ$; $\theta_{S4} = 29.0^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

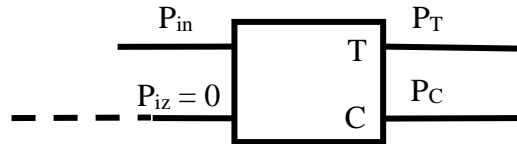
Subject no. 45

1. $z = 1.250 - j \cdot 0.835$; $Y = 1 / 50\Omega / (1.250 - j \cdot 0.835) = 0.0111S + j \cdot (0.0074)S$; $\Gamma = (z-1)/(z+1) = (1.250 - j \cdot 0.835 - 1)/(1.250 - j \cdot 0.835 + 1) = 0.219 + j \cdot (-0.290) = 0.363 \angle -53.0^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.129$, $Z_{0E} = 56.916\Omega$, $Z_{0O} = 43.925\Omega$

b) $P_c = 70.5\mu W = -11.518\text{dBm}$; $P_{in} = P_c + C = -11.518\text{dBm} + 17.8\text{dB} = 6.282\text{dBm} = 4.248 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [\text{mW}] = P_{in} - P_c - P_{iz} = 4.248\text{mW} - 0.0705\text{mW} - 0 = 4.178 \text{ mW} = 6.209 \text{ dBm}$



3. The shunt RC load with $R = 33\Omega$ and $C = 0.557\text{pF}$ has $Z_L = 17.40\Omega + j \cdot (-16.48)\Omega$ at 8.2GHz

$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$ (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 4/8) = 0.000$; $\cot(\beta l) = \infty$; $Z_{in} = 23.96\Omega \angle -43.4^\circ = 17.40\Omega + j \cdot (-16.48)\Omega$;

b) If the line becomes open-circuited $Z_L = \infty$; $Z_{in} = j \cdot (0.00)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 15\text{dB} + 17\text{dB} + 17\text{dB} = 49\text{dB}$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$F_1 = 2.01\text{dB} = 1.589$, $G_1 = 15\text{dB} = 31.623$, $F_2 = 2.48\text{dB} = 1.770$, $G_2 = 17\text{dB} = 50.119$, $F_3 = 2.81\text{dB} = 1.910$,
 $F = 1.589 + (1.770 - 1)/31.623 + (1.910 - 1)/31.623/50.119 = 1.613 = 2.078\text{dB}$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ |
|----|-----------------------------|------------|-------|-------|
| T1 | $-0.437 + j \cdot (0.183)$ | 0.474 | 0.801 | 0.698 |
| T2 | $-0.226 + j \cdot (-0.107)$ | 0.250 | 0.931 | 0.947 |

b) $\mu (T1) < \mu (T2)$ so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.061$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 23.364 = 13.685 \text{ dB}$ (L9/2023, S75)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.118$, $\arg(S_{22}^*) = 157.2^\circ$; $\theta_{S1} = 149.8^\circ$; $\text{Im}(y_S) = -0.238$; **or**
 $\theta_{S2} = 53.0^\circ$; $\text{Im}(y_S) = 0.238$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.622$, $\arg(S_{11}^*) = 137.4^\circ$; $\theta_{S3} = 175.5^\circ$; $\text{Im}(y_S) = -1.589$; **or**
 $\theta_{S4} = 47.1^\circ$; $\text{Im}(y_S) = 1.589$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.238) + (-1.589) = -1.826$; $\theta_{S1} = 149.8^\circ$; $\theta_{P1} = 118.7^\circ$; $\theta_{S3} = 175.5^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.238) + (1.589) = 1.351$; $\theta_{S1} = 149.8^\circ$; $\theta_{P2} = 53.5^\circ$; $\theta_{S4} = 47.1^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.238) + (-1.589) = -1.351$; $\theta_{S2} = 53.0^\circ$; $\theta_{P3} = 126.5^\circ$; $\theta_{S3} = 175.5^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.238) + (1.589) = 1.826$; $\theta_{S2} = 53.0^\circ$; $\theta_{P4} = 61.3^\circ$; $\theta_{S4} = 47.1^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

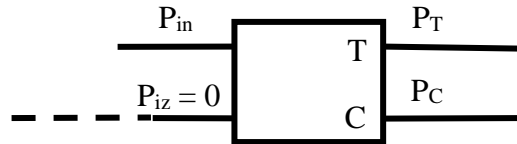
Subject no. 46

1. $z = 1.005 - j \cdot 1.190$; $Y = 1 / 50\Omega / (1.005 - j \cdot 1.190) = 0.0083S + j \cdot (0.0098)S$; $\Gamma = (z-1)/(z+1) = (1.005 - j \cdot 1.190 - 1)/(1.005 - j \cdot 1.190 + 1) = 0.262 + j \cdot (-0.438) = 0.510 \angle -59.1^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.072$, $Z_{0E} = 53.719\Omega$, $Z_{0O} = 46.539\Omega$

b) $P_c = 72.0\mu W = -11.427\text{dBm}$; $P_{in} = P_c + C = -11.427\text{dBm} + 22.9\text{dB} = 11.473\text{dBm} = 14.039 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [\text{mW}] = P_{in} - P_c - P_{iz} = 14.039\text{mW} - 0.0720\text{mW} - 0 = 13.967 \text{ mW} = 11.451 \text{ dBm}$



3. The series RC load with $R = 73\Omega$ and $C = 0.363\text{pF}$ has $Z_L = 73.00\Omega + j \cdot (-49.26)\Omega$ at 8.9GHz

$$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$$
 (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 4/6) = 1.732$; $\cot(\beta l) = 0.577$; $Z_{in} = 39.95\Omega \angle 9.6^\circ = 39.40\Omega + j \cdot (6.65)\Omega$;

b) If the line becomes open-circuited $Z_L = \infty$; $Z_{in} = j \cdot (-43.30)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 12\text{dB} + 10\text{dB} + 17\text{dB} = 39\text{dB}$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$F_1 = 2.03\text{dB} = 1.596$, $G_1 = 12\text{dB} = 15.849$, $F_2 = 2.08\text{dB} = 1.614$, $G_2 = 10\text{dB} = 10.000$, $F_3 = 2.29\text{dB} = 1.694$,
 $F = 1.596 + (1.614 - 1)/15.849 + (1.694 - 1)/15.849/10.000 = 1.639 = 2.146\text{dB}$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ' |
|----|-----------------------------|------------|-------|--------|
| T1 | $-0.457 + j \cdot (0.125)$ | 0.474 | 0.771 | 0.824 |
| T2 | $-0.232 + j \cdot (-0.139)$ | 0.270 | 0.898 | 0.951 |

b) μ' (T1) < μ' (T2) so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.080$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 23.773 = 13.761 \text{ dB}$ (L9/2023, S75)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.143$, $\arg(S_{22}^*) = 146.7^\circ$; $\theta_{S1} = 155.8^\circ$; $\text{Im}(y_S) = -0.289$; **or**
 $\theta_{S2} = 57.6^\circ$; $\text{Im}(y_S) = 0.289$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.631$, $\arg(S_{11}^*) = 131.7^\circ$; $\theta_{S3} = 178.7^\circ$; $\text{Im}(y_S) = -1.627$; **or**
 $\theta_{S4} = 49.6^\circ$; $\text{Im}(y_S) = 1.627$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.289) + (-1.627) = -1.916$; $\theta_{S1} = 155.8^\circ$; $\theta_{P1} = 117.6^\circ$; $\theta_{S3} = 178.7^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.289) + (1.627) = 1.338$; $\theta_{S1} = 155.8^\circ$; $\theta_{P2} = 53.2^\circ$; $\theta_{S4} = 49.6^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.289) + (-1.627) = -1.338$; $\theta_{S2} = 57.6^\circ$; $\theta_{P3} = 126.8^\circ$; $\theta_{S3} = 178.7^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.289) + (1.627) = 1.916$; $\theta_{S2} = 57.6^\circ$; $\theta_{P4} = 62.4^\circ$; $\theta_{S4} = 49.6^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

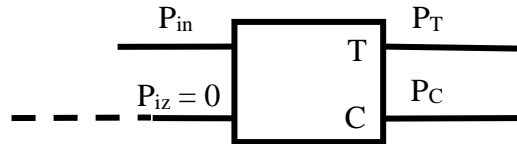
Subject no. 47

1. $z = 1.260 + j \cdot 0.925$; $Y = 1 / 50\Omega / (1.260 + j \cdot 0.925) = 0.0103S + j \cdot (-0.0076)S$; $\Gamma = (z-1)/(z+1) = (1.260 + j \cdot 0.925 - 1)/(1.260 + j \cdot 0.925 + 1) = 0.242 + j \cdot (0.310) = 0.393 \angle 52.0^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.122$, $Z_{0E} = 56.500\Omega$, $Z_{0O} = 44.248\Omega$

b) $P_c = 110.5\mu W = -9.566\text{dBm}$; $P_{in} = P_c + C = -9.566\text{dBm} + 18.3\text{dB} = 8.734\text{dBm} = 7.471 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [\text{mW}] = P_{in} - P_c - P_{iz} = 7.471\text{mW} - 0.1105\text{mW} - 0 = 7.360 \text{ mW} = 8.669 \text{ dBm}$



3. The series RL load with $R = 40\Omega$ and $L = 1.230\text{nH}$ has $Z_L = 40.00\Omega + j \cdot (62.60)\Omega$ at 8.1GHz

$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$ (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 2/5) = -0.727$; $\cot(\beta l) = -1.376$; $Z_{in} = 23.97\Omega \angle 50.2^\circ = 15.34\Omega + j \cdot (18.43)\Omega$;

b) If the line becomes open-circuited $Z_L = \infty$; $Z_{in} = j \cdot (68.82)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 18\text{dB} + 12\text{dB} + 14\text{dB} = 44\text{dB}$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$F_1 = 2.45\text{dB} = 1.758$, $G_1 = 18\text{dB} = 63.096$, $F_2 = 2.42\text{dB} = 1.746$, $G_2 = 12\text{dB} = 15.849$, $F_3 = 2.50\text{dB} = 1.778$,
 $F = 1.758 + (1.746 - 1)/63.096 + (1.778 - 1)/63.096/15.849 = 1.771 = 2.481\text{dB}$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ |
|----|----------------------------|------------|-------|-------|
| T1 | $-0.009 + j \cdot (0.493)$ | 0.493 | 0.950 | 0.913 |
| T2 | $-0.113 + j \cdot (0.007)$ | 0.113 | 1.215 | 1.159 |

b) $\mu (T1) < \mu (T2)$ so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.071$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 20.236 = 13.061 \text{ dB}$ (L9/2023, S75); However $K = 1.215 > 1$ so maximum gain is $MAG = |S_{21}| / |S_{12}| \cdot [K - \sqrt{K^2 - 1}] = 10.622 = 10.262 \text{ dB}$ (L9/2023, S76)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.179$, $\arg(S_{22}^*) = -55.1^\circ$; $\theta_{S1} = 77.7^\circ$; $\text{Im}(y_S) = -0.364$; **or**
 $\theta_{S2} = 157.4^\circ$; $\text{Im}(y_S) = 0.364$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.571$, $\arg(S_{11}^*) = -177.0^\circ$; $\theta_{S3} = 150.9^\circ$; $\text{Im}(y_S) = -1.391$; **or**
 $\theta_{S4} = 26.1^\circ$; $\text{Im}(y_S) = 1.391$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.364) + (-1.391) = -1.755$; $\theta_{S1} = 77.7^\circ$; $\theta_{P1} = 119.7^\circ$; $\theta_{S3} = 150.9^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.364) + (1.391) = 1.027$; $\theta_{S1} = 77.7^\circ$; $\theta_{P2} = 45.8^\circ$; $\theta_{S4} = 26.1^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.364) + (-1.391) = -1.027$; $\theta_{S2} = 157.4^\circ$; $\theta_{P3} = 134.2^\circ$; $\theta_{S3} = 150.9^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.364) + (1.391) = 1.755$; $\theta_{S2} = 157.4^\circ$; $\theta_{P4} = 60.3^\circ$; $\theta_{S4} = 26.1^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

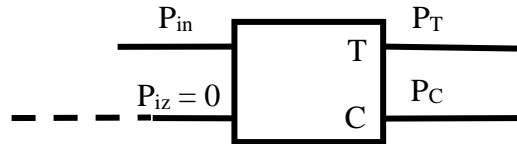
Subject no. 48

1. $z = 1.165 - j \cdot 0.725$; $Y = 1 / 50\Omega / (1.165 - j \cdot 0.725) = 0.0124S + j \cdot (0.0077)S$; $\Gamma = (z-1)/(z+1) = (1.165 - j \cdot 0.725 - 1)/(1.165 - j \cdot 0.725 + 1) = 0.169 + j \cdot (-0.278) = 0.326 \angle -58.7^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.079$, $Z_{0E} = 54.143\Omega$, $Z_{0O} = 46.174\Omega$

b) $P_c = 71.0\mu W = -11.487\text{dBm}$; $P_{in} = P_c + C = -11.487\text{dBm} + 22.0\text{dB} = 10.513\text{dBm} = 11.253 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [\text{mW}] = P_{in} - P_c - P_{iz} = 11.253\text{mW} - 0.0710\text{mW} - 0 = 11.182 \text{ mW} = 10.485 \text{ dBm}$



3. The series RC load with $R = 56\Omega$ and $C = 0.435\text{pF}$ has $Z_L = 56.00\Omega + j \cdot (-51.53)\Omega$ at 7.1GHz

$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$ (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 2/8) = \infty$; $\cot(\beta l) = 0.000$; $Z_{in} = 16.10\Omega \angle 42.6^\circ = 11.84\Omega + j \cdot (10.90)\Omega$;

b) If the line becomes open-circuited $Z_L = \infty$; $Z_{in} = j \cdot (0.00)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 12\text{dB} + 11\text{dB} + 19\text{dB} = 42\text{dB}$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$F_1 = 2.46\text{dB} = 1.762$, $G_1 = 12\text{dB} = 15.849$, $F_2 = 2.64\text{dB} = 1.837$, $G_2 = 11\text{dB} = 12.589$, $F_3 = 2.93\text{dB} = 1.963$,
 $F = 1.762 + (1.837 - 1)/15.849 + (1.963 - 1)/15.849/12.589 = 1.820 = 2.600\text{dB}$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ' |
|----|-----------------------------|------------|-------|--------|
| T1 | $-0.472 + j \cdot (-0.107)$ | 0.484 | 0.658 | 0.764 |
| T2 | $-0.221 + j \cdot (-0.248)$ | 0.333 | 0.793 | 0.905 |

b) $\mu' (T1) < \mu' (T2)$ so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.178$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 26.981 = 14.311 \text{ dB}$ (L9/2023, S75)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.232$, $\arg(S_{22}^*) = 119.1^\circ$; $\theta_{S1} = 172.2^\circ$; $\text{Im}(y_S) = -0.477$; **or**
 $\theta_{S2} = 68.8^\circ$; $\text{Im}(y_S) = 0.477$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.682$, $\arg(S_{11}^*) = 112.0^\circ$; $\theta_{S3} = 10.5^\circ$; $\text{Im}(y_S) = -1.865$; **or**
 $\theta_{S4} = 57.5^\circ$; $\text{Im}(y_S) = 1.865$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.477) + (-1.865) = -2.342$; $\theta_{S1} = 172.2^\circ$; $\theta_{P1} = 113.1^\circ$; $\theta_{S3} = 10.5^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.477) + (1.865) = 1.388$; $\theta_{S1} = 172.2^\circ$; $\theta_{P2} = 54.2^\circ$; $\theta_{S4} = 57.5^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.477) + (-1.865) = -1.388$; $\theta_{S2} = 68.8^\circ$; $\theta_{P3} = 125.8^\circ$; $\theta_{S3} = 10.5^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.477) + (1.865) = 2.342$; $\theta_{S2} = 68.8^\circ$; $\theta_{P4} = 66.9^\circ$; $\theta_{S4} = 57.5^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

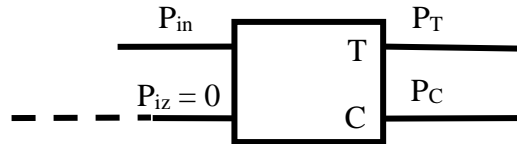
Subject no. 49

1. $z = 0.955 - j \cdot 0.850$; $Y = 1 / 50\Omega / (0.955 - j \cdot 0.850) = 0.0117S + j \cdot (0.0104)S$; $\Gamma = (z-1)/(z+1) = (0.955 - j \cdot 0.850 - 1)/(0.955 - j \cdot 0.850 + 1) = 0.140 + j \cdot (-0.374) = 0.399 \angle -69.5^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.129$, $Z_{0E} = 56.916\Omega$, $Z_{0O} = 43.925\Omega$

b) $P_c = 91.5\mu W = -10.386\text{dBm}$; $P_{in} = P_c + C = -10.386\text{dBm} + 17.8\text{dB} = 7.414\text{dBm} = 5.513 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [\text{mW}] = P_{in} - P_c - P_{iz} = 5.513\text{mW} - 0.0915\text{mW} - 0 = 5.422 \text{ mW} = 7.342 \text{ dBm}$



3. The series RC load with $R = 73\Omega$ and $C = 0.337\text{pF}$ has $Z_L = 73.00\Omega + j \cdot (-54.92)\Omega$ at 8.6GHz

$$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$$
 (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 3/6) = 0.000$; $\cot(\beta l) = \infty$; $Z_{in} = 91.35\Omega \angle -37.0^\circ = 73.00\Omega + j \cdot (-54.92)\Omega$;

b) If the line becomes open-circuited $Z_L = \infty$; $Z_{in} = j \cdot (0.00)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 16\text{dB} + 10\text{dB} + 11\text{dB} = 37\text{dB}$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$F_1 = 2.75\text{dB} = 1.884$, $G_1 = 16\text{dB} = 39.811$, $F_2 = 2.61\text{dB} = 1.824$, $G_2 = 10\text{dB} = 10.000$, $F_3 = 2.74\text{dB} = 1.879$,
 $F = 1.884 + (1.824 - 1)/39.811 + (1.879 - 1)/39.811/10.000 = 1.907 = 2.802\text{dB}$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ |
|----|-----------------------------|------------|-------|-------|
| T1 | $-0.458 + j \cdot (-0.179)$ | 0.492 | 0.622 | 0.494 |
| T2 | $-0.208 + j \cdot (-0.284)$ | 0.352 | 0.763 | 0.810 |

b) $\mu (T1) < \mu (T2)$ so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.225$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 28.300 = 14.518 \text{ dB}$ (L9/2023, S75)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.260$, $\arg(S_{22}^*) = 111.8^\circ$; $\theta_{S1} = 176.6^\circ$; $\text{Im}(y_S) = -0.539$; **or**
 $\theta_{S2} = 71.6^\circ$; $\text{Im}(y_S) = 0.539$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.700$, $\arg(S_{11}^*) = 106.0^\circ$; $\theta_{S3} = 14.2^\circ$; $\text{Im}(y_S) = -1.960$; **or**
 $\theta_{S4} = 59.8^\circ$; $\text{Im}(y_S) = 1.960$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.539) + (-1.960) = -2.499$; $\theta_{S1} = 176.6^\circ$; $\theta_{P1} = 111.8^\circ$; $\theta_{S3} = 14.2^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.539) + (1.960) = 1.422$; $\theta_{S1} = 176.6^\circ$; $\theta_{P2} = 54.9^\circ$; $\theta_{S4} = 59.8^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.539) + (-1.960) = -1.422$; $\theta_{S2} = 71.6^\circ$; $\theta_{P3} = 125.1^\circ$; $\theta_{S3} = 14.2^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.539) + (1.960) = 2.499$; $\theta_{S2} = 71.6^\circ$; $\theta_{P4} = 68.2^\circ$; $\theta_{S4} = 59.8^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

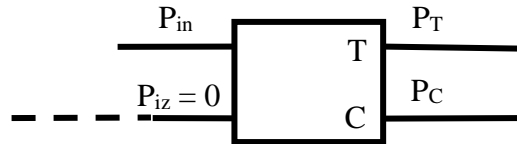
Subject no. 50

1. $z = 1.250 - j \cdot 0.750$; $Y = 1 / 50\Omega / (1.250 - j \cdot 0.750) = 0.0118S + j \cdot (0.0071)S$; $\Gamma = (z-1)/(z+1) = (1.250 - j \cdot 0.750 - 1)/(1.250 - j \cdot 0.750 + 1) = 0.200 + j \cdot (-0.267) = 0.333 \angle -53.1^\circ$, plot point in complex plane either with rectangular coordinates or polar coordinates

2. a) $\beta = 10^{-C/20} = 0.077$, $Z_{0E} = 53.996\Omega$, $Z_{0O} = 46.300\Omega$

b) $P_c = 54.0\mu W = -12.676\text{dBm}$; $P_{in} = P_c + C = -12.676\text{dBm} + 22.3\text{dB} = 9.624\text{dBm} = 9.171 \text{ mW}$;

c) Lossless/Ideal coupler: $P_T [\text{mW}] = P_{in} - P_c - P_{iz} = 9.171\text{mW} - 0.0540\text{mW} - 0 = 9.117 \text{ mW} = 9.598 \text{ dBm}$



3. The shunt RL load with $R = 35\Omega$ and $L = 0.638\text{nH}$ has $Z_L = 14.86\Omega + j \cdot (17.30)\Omega$ at 7.5GHz

$Z_{in} = Z_0 \cdot \frac{Z_L + j \cdot Z_0 \cdot \tan(\beta l)}{Z_0 + j \cdot Z_L \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 + j \cdot \tan(\beta l)}{1 + j \cdot Z_L/Z_0 \cdot \tan(\beta l)} = Z_0 \cdot \frac{Z_L/Z_0 \cdot \cot(\beta l) + j}{\cot(\beta l) + j \cdot Z_L/Z_0}$ (use the appropriate formula if tan or cot are ∞);

a) $\tan(\beta l) = \tan(2 \cdot \pi \cdot 5/8) = 1.000$; $\cot(\beta l) = 1.000$; $Z_{in} = 95.94\Omega \angle 53.1^\circ = 57.59\Omega + j \cdot (76.73)\Omega$;

b) If the line becomes short-circuited $Z_L = 0$; $Z_{in} = j \cdot (50.00)\Omega$

4. a) $G = G_1 + G_2 + G_3 = 16\text{dB} + 10\text{dB} + 11\text{dB} = 37\text{dB}$

b) Friis formula, $F = F_1 + (F_2 - 1)/G_1 + (F_3 - 1)/G_1/G_2$

$F_1 = 2.27\text{dB} = 1.687$, $G_1 = 16\text{dB} = 39.811$, $F_2 = 2.99\text{dB} = 1.991$, $G_2 = 10\text{dB} = 10.000$, $F_3 = 2.52\text{dB} = 1.786$,
 $F = 1.687 + (1.991 - 1)/39.811 + (1.786 - 1)/39.811/10.000 = 1.713 = 2.339\text{dB}$

5. a) Must compute either μ or μ' (as requested!) (L8/2023, S96 or 97);

| | Δ | $ \Delta $ | K | μ' |
|----|----------------------------|------------|-------|--------|
| T1 | $-0.134 + j \cdot (0.463)$ | 0.482 | 0.917 | 0.928 |
| T2 | $-0.140 + j \cdot (0.000)$ | 0.140 | 1.156 | 1.069 |

b) $\mu'(T1) < \mu'(T2)$ so the transistor T2 has better stability

c) For T1: $U = |S_{11}| \cdot |S_{12}| \cdot |S_{21}| \cdot |S_{22}| / (1 - |S_{11}|^2) / (1 - |S_{22}|^2) = 0.052$ (L9/2023, S101)

d) For T2: $MSG = |S_{21}| / |S_{12}| = 20.964 = 13.215 \text{ dB}$ (L9/2023, S75); However $K = 1.156 > 1$ so maximum gain is $MAG = |S_{21}| / |S_{12}| \cdot [K - \sqrt{K^2 - 1}] = 12.066 = 10.816 \text{ dB}$ (L9/2023, S76)

e) Match is performed from S_{22}^* for T1 to S_{11}^* for T2, Complex calculus from L7/2023, S84-88

S_{22}^* for T1: 2 solutions for the match, $|S_{22}| = 0.128$, $\arg(S_{22}^*) = -76.1^\circ$; $\theta_{S1} = 86.7^\circ$; $\text{Im}(y_S) = -0.258$; **or**
 $\theta_{S2} = 169.4^\circ$; $\text{Im}(y_S) = 0.258$

S_{11}^* for T2: 2 solutions for the match, $|S_{11}| = 0.576$, $\arg(S_{11}^*) = 173.0^\circ$; $\theta_{S3} = 156.1^\circ$; $\text{Im}(y_S) = -1.409$; **or**
 $\theta_{S4} = 30.9^\circ$; $\text{Im}(y_S) = 1.409$

Combining the two schematics (L13/2023, S96), all lines with $Z_0 = 50\Omega$:

Solution 1: $\text{Im}(y_S) = (-0.258) + (-1.409) = -1.667$; $\theta_{S1} = 86.7^\circ$; $\theta_{P1} = 121.0^\circ$; $\theta_{S3} = 156.1^\circ$; **or**

Solution 2: $\text{Im}(y_S) = (-0.258) + (1.409) = 1.151$; $\theta_{S1} = 86.7^\circ$; $\theta_{P2} = 49.0^\circ$; $\theta_{S4} = 30.9^\circ$; **or**

Solution 3: $\text{Im}(y_S) = (0.258) + (-1.409) = -1.151$; $\theta_{S2} = 169.4^\circ$; $\theta_{P3} = 131.0^\circ$; $\theta_{S3} = 156.1^\circ$; **or**

Solution 4: $\text{Im}(y_S) = (0.258) + (1.409) = 1.667$; $\theta_{S2} = 169.4^\circ$; $\theta_{P4} = 59.0^\circ$; $\theta_{S4} = 30.9^\circ$;

f) Schematic drawing, "T" shaped, the combined shunt stub $\theta_{P1}/\theta_{P2}/\theta_{P3}/\theta_{P4}$ must be between the two series lines

